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ILS GLIDE SLOPE PERFORMANCE PREDICTION MULTIPATH SCATTERING

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16. Abstract

A mathematical model has been developed which predicts the performance of ILS glide slope systems subject to multipath scattering and the effects of irregular terrain contours. The model is discussed in detail and then applied to a test case for purposes of illustration. A complete listing of all computer programs has been appended to the report. A users' manual has been prepared under separate cover.

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PREFACE

The work presented in this report has been sponsored by the Federal Aviation Administration as part of a program to provide navigational aids for the safe landing of aircraft. The program has been concerned with the application of instrument landing aids, and in particular the development of models to predict the performance of localizer and glide slope antenna systems. The present report is concerned with the glide slope portion of instrument landing system, and the effects of airport topography on its performance.

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CONTENTS

Page

Section

1.	INTRODUCTION	1
2.	BASIC THEORY	1
	Part A Part B	1 17
3.	NUMERICAL RESULTS	19
	ILLUSTRATIONS	
Figure		Page
Figure 1.	TYPICAL GEOMETRY FOR MULTIPATH SCATTERING	Page 2
	TYPICAL GEOMETRY FOR MULTIPATH SCATTERING REPRESENTATION OF RECTANGULAR FACET	
1.	REPRESENTATION OF RECTANGULAR FACET TYPICAL PROFILE OF ONE-DIMENSIONAL TERRAIN	2
1. 2. 3.	REPRESENTATION OF RECTANGULAR FACET TYPICAL PROFILE OF ONE-DIMENSIONAL TERRAIN VARIATION	2
1.	REPRESENTATION OF RECTANGULAR FACET TYPICAL PROFILE OF ONE-DIMENSIONAL TERRAIN	2 9
1. 2. 3.	REPRESENTATION OF RECTANGULAR FACET TYPICAL PROFILE OF ONE-DIMENSIONAL TERRAIN VARIATION	2 9
1. 2. 3.	REPRESENTATION OF RECTANGULAR FACET TYPICAL PROFILE OF ONE-DIMENSIONAL TERRAIN VARIATION OVERVIEW OF A MULTIPATH CONFIGURATION	2 9 18 24

1. INTRODUCTION

A mathematical model has been developed for predicting the performance of image-type glide slope arrays. The basic theory is developed in the following section of this report. In Part A of that section a model of the glide slope multipath problem is developed while in Part B the techniques for dealing with irregular terrain profiles are outlined. In the final section of the report some illustrative numerical results are presented.

2. BASIC THEORY PART A

The presence in the airport environment of such large manmade structures as aircraft hangars as well as such natural terrain features as hillsides can lead to glide slope course derogation through multipath scattering. We begin our treatment of glide slope siting problems by developing a model for predicting the amount of such multipath derogation. A typical situation is depicted in Figure 1.

The touchdown point on the centerline of the runway opposite the glide slope array is chosen as the origin of coordinates 0. The z-axis is chosen to be the vertical axis passing through the origin 0, while the x-and y-axes are parallel to and perpendicular to the centerline, respectively. The positive z-axis points out of the page (Figure 1). The ground plane is assumed to be perfectly conducting. Consequently, any deviations from nominal glide slope performance are attributable to the multipath scattering produced by the various structures (both natural and man-made). For simplicity, we will assume that all such structures are perfectly conducting.

In the situation depicted in Figure 1, the scatterer (perhaps a small hill) is illuminated by the glide slope array and scatters

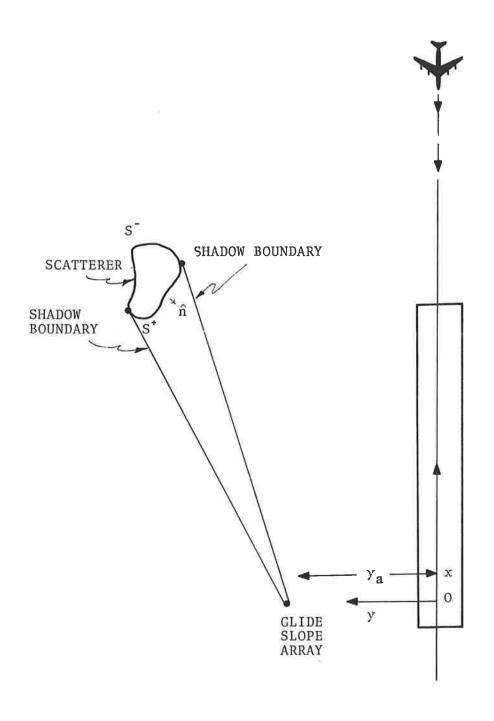


FIGURE 1. TYPICAL GEOMETRY FOR MULTIPATH SCATTERING

some of this signal back into the path of the approaching aircraft. We now proceed to calculate mathematical expressions for these scattered fields at the receiver. For convenience, we will work with magnetic field intensities.

For a perfect conductor, the surface current density K is given by:

$$K = \hat{n} \times H$$
 (1)

where n is the local unit normal vector pointing out of the scattering surface and H is the total magnetic field,

$$\overrightarrow{H} = \overrightarrow{H}_{1} + \overrightarrow{H}_{S}, \tag{2}$$

where H_i is the incident and H_s the scattered field on the surface of the scattering structure. In terms of the surface current density K, the scattered field H_s at the receiver is given by the following surface integral:

$$\overrightarrow{H}_{S}(\overrightarrow{r}_{1}) = \frac{1}{4\pi} \int_{S} \left(\overrightarrow{K}(\overrightarrow{r}) x \overrightarrow{\nabla} G(\overrightarrow{r}_{1}, \overrightarrow{r}) \right) ds$$
 (3)

In Equation (3), the vectors \vec{r}_1 and \vec{r} denote, respectively, the position vector of the receiver and the position vector of some arbitrary source point on the surface of the scatterer relative to the origin of coordinates 0. The two-point Green's function $G(\vec{r}_1,\vec{r})$ is given by:

$$G(\overline{r_1},\overline{r}) = \underbrace{e^{ik|\overline{r_1}-\overline{r}|}}_{|\overline{r_1}-\overline{r}|}$$
(4)

where k = 2 $\pi/\,\lambda$ and λ is the wavelength of the incident radiation. The integral is taken over the surface S of the scatterer.

We will adopt here the single scattering, physical optics approximation for the current density K. Specifically, we will assume that on those portions of the scattering surface not directly illuminated by the glide slope array, K is identically zero and

that on the directly illuminated portions of the scattering surface, K is proportional to twice the tangential component of the incident magnetic field:

$$\vec{K} \equiv 0 \text{ on } S_{\underline{}}$$
 (5)

$$\overrightarrow{K} \equiv 0 \text{ on } S_{\underline{}}$$

$$\overrightarrow{K} = 2 \left(\stackrel{\wedge}{n} \times \overrightarrow{H}_{\underline{i}} \right) \text{ on } S_{\underline{+}}$$
(5)

where S_{+} and S_{-} denote, respectively, the illuminated and unilluminated portions of the scattering surface S (see Figure 1). For the current distribution defined by Equations (5) and (6), Equation (3) becomes

$$\overrightarrow{H}_{S}(\mathbf{r}_{1}) = \frac{1}{2\pi} \int \left[\widehat{\mathbf{n}}(\overrightarrow{\mathbf{r}}) x \overrightarrow{\mathbf{H}}_{1}(\overrightarrow{\mathbf{r}}) \right] x \overrightarrow{\nabla} G(\overrightarrow{\mathbf{r}}_{1}, \overrightarrow{\mathbf{r}}) ds.$$
(7)

The physical optics scattering model represented by equation (7) is based upon the existence of a sharp shadow boundary on the surface of the scattering structure. This assumption ignores such phenomena multiple reflections, surface waves, and diffraction but should provide reasonably accurate results as long as the surface features of the scattering structure do not vary greatly over dimensions which are small compared to the wavelength λ (about 3 feet at glide slope frequencies).

In principle, the solution to the problem at hand requires only that the magnet field \mathbf{H}_{i} be determined. For simplicity, we will assume that the glide slope array is made up of electrically short dipoles. With this assumption, we in effect approximate by a cosine distribution the actual azimuthal pattern of the half-wave dipoles which make up glide slope arrays. This approximation should not be too restrictive and it does considerably expedite our calculations. Let $(0, y_a, h)$ denote the x, y, and z coordinates respectively of a typical dipole in the array and let (x, y, z) denote the coordinates of a point \vec{r} at which we wish to know the magnetic field intensity Ha produced by the dipole at (0, ya, h). In our earlier

glide slope performance prediction report,* it is shown that $\overline{H}_a(r)$ is given by

$$\overline{H_a(r)} = \frac{ikJ_o}{4\pi} \frac{e^{ikD_1}}{D_1^2} \left[xe_z - (z-h)e_x\right]$$
(8)

where e_x^{λ} , e_y^{λ} , and e_z^{λ} denote unit vectors in the x, y, and z directions respectively, J_o is a parameter which measures relative phase and amplitude, and D_1 is the distance from the antenna to the observation point r:

$$D_{1} = \left[x^{2} + (y - y_{a})^{2} + (z - h)^{2}\right]^{\frac{1}{2}}$$
(9)

To take into account reflections from the ground plane (z = 0) we use simple image theory to obtain the image field $H_2(\vec{r})$:

$$\frac{\mathbf{H}_{a}^{\star}(\mathbf{r})}{\mathbf{H}_{a}^{\star}(\mathbf{r})} = \frac{i k J_{o}}{4\pi} = \frac{e^{i k D_{2}}}{D_{2}^{2}} \left[x e_{z}^{\bullet} - (z+h) e_{x}^{\bullet} \right]$$
(10)

where D_2 is the distance from the image of the transmitting dipole to the field point \overline{r} :

$$D_{2} = \left[x^{2} + (y-y_{a})^{2} + (z+h)^{2}\right]^{\frac{1}{2}}$$
(11)

The total field intensity $H_{\dot{1}}$ at $\dot{\vec{r}}$ is just the sum of the direct $(\dot{\vec{H}}_a)$ and ground reflected $(\dot{\vec{H}}_a^*)$ signals:

$$\overline{H}_{i}(r) = \overline{H}_{a}(r) + \overline{H}_{a}(r)$$
(12)

Note that on the ground plane (z = 0), the z component of H_i vanishes identically.

^{*} ILS Glide Slope Performance Predictions, Vol. B, FAA-RD-74-157.B, S. Morin, D. Newsom, D. Kahn, L. Jordan, September 1974.

The field H_i determined by Equations (8), (10), and (12) can now be substituted into Equation (7) and the integral can be performed over the illuminated surface of the scattering structure to give the scattered field at the receiver location r_1 . Actually, only the z component of the scattered field need be calculated since the receiving antenna responds primarily to the horizontal component of electric field which is proportional to the z component of the magnetic field. It should be noted that for a complete solution, the contribution of the image of the scattering structure must also be calculated. We will assume that the receiver is always in the far field of the scattering structure so that the following asymptotic form of the gradient of the Green's function is applicable:

$$\nabla G(\vec{r}_1, \vec{r}) \approx -ik (\vec{r}_1 - \vec{r}) \frac{e^{ik|\vec{r}_1 - \vec{r}|}}{|\vec{r}_1 - \vec{r}|^2}$$
 (13)

Substituting Euqations (8), (10), (12) and (13) into Equation (7) and taking the z-component of the resulting vector equation, we obtain the following expressions for $H_{\rm SZ}({\bf r}_1)$, the z-component of scattered magnetic field at the receiver point ${\bf r}_1$:

$$H_{sz}(\vec{r}_1) = \frac{k^2 J_0}{8\pi^2} \qquad \left(I_1 + I_2 + I_3 + I_4\right)$$
 (14)

where

$$I_{1} = \int_{S_{+}}^{F_{1}} \frac{F_{1}}{D_{1}^{2}R_{1}^{2}} e^{ik(R_{1}+D_{1})} ds$$
 (15)

$$I_2 = -\int_{S_+} \frac{F_2}{D_2^2 R_1^2} e^{ik(R_1 + D_2)} ds$$
 (16)

$$I_3 = -\int_{S_+} \frac{F_1}{D_1^2 R_2^2} e^{ik(R_2 + D_1)} ds$$
 (17)

$$I_4 = \int_{S_+} \frac{F_2}{D_2^2 R_2^2} = e^{ik(R_2 + D_2)} ds$$
 (18)

$$D_{1} = \left[x^{2} + (y-y_{a})^{2} + (z-h)^{2}\right]^{\frac{1}{2}}$$
 (19)

$$D_2 = \left[x^2 + (y - y_a)^2 + (z + h)^2 \right]^{\frac{1}{2}}$$
 (20)

$$R_{1} = \left[(X_{1} - x)^{2} + (Y_{1} - y)^{2} + (Z_{1} - z)^{2} \right]^{\frac{1}{2}}$$
 (21)

$$R_2 = \left[(X_1 - x)^2 + (Y_1 - y)^2 + (Z_1 + z)^2 \right]^{\frac{1}{2}}$$
 (22)

$$F_1 = n_2 x (Y_1 - y) + (n_1 x + n_3 (z - h)) (X_1 - x),$$
 (23)

$$F_2 = n_2 x (Y_1 - y) + (n_1 x + n_3 (z + h)) (X_1 - x)$$
 (24)

In Equations (14) through (24), the coordinates (X_1, Y_1, Z_1) denote the coordinates of the aircraft, (x, y, z) the coordinates of some source point on the surface of the scattering structure and (n_1, n_2, n_3) the components of the unit outward normal to the surface of the scatterer at the point (x, y, z). The integrals I_3 and I_4 represent the contributions of the image of the scattering structure in the ground plane z = 0. Note the following relationships among the four integrals:

$$I_1 (X_1, Y_1, Z_1) = -I_3 (X_1, Y_1, -Z_1)$$

 $I_2 (X_1, Y_1, Z_1) = -I_4 (X_1, Y_1, -Z_1)$ (25)

Consequently, when $Z_1 = 0$, $H_{SZ}(r_1) \equiv 0$ as it should be.

The direct numerical integration of the integrals \mathbf{I}_1 , \mathbf{I}_2 , \mathbf{I}_3 , and I_{Δ} for an arbitrary scattering structure would in general require an inordinate amount of computer time. However, certain simplifying assumptions can be made which considerably expedite the evaluation of these integrals. Specifically, we will assume that any scattering surface can be adequately represented by a series of interconnecting plane facets. By definition, the unit normal vector n will be constant over each facet. The exact size, shape, and orientation of these facets will depend upon the surface characteristics of the particular scatterer. In turn, each facet can be broken up into a series of interconnected rectangular pieces. If these rectangular pieces are made small enough, the integrals I1, ${\rm I}_2$, ${\rm I}_3$ and ${\rm I}_4$ can be evaluated analytically for each piece and then the results summed over the whole surface of the scatterer to give $H_{s.z}$. To illustrate these procedures, we will now evaluate the integral I_1 for an arbitrarily oriented rectangular plate. In the course of this evaluation, we will discuss the size restrictions on such plates.

In Figure 2, a rectangular facet typical of those making up the surface of some scatterer has been drawn. We will evaluate the integral I_1 of Equation (15) for such a facet in the Fraunhofer approximation. The unit vectors $\stackrel{\wedge}{\eta}$ and $\stackrel{\wedge}{\xi}$ lie in the plane of the facet, are orthonormal, and define the direction of the outward normal $\stackrel{\wedge}{\eta}$:

$$n = n \times \xi$$

The center point of the facet (x_0, y_0, z_0) will be used as a local origin of coordinates for the integrations which are to be performed. The integral I_1 is given by

$$I_{1} = \int_{S_{1}} \frac{F_{1}}{D_{1}^{2} R_{1}^{2}} e^{ik(D_{1}+R_{1})} ds$$
 (26)

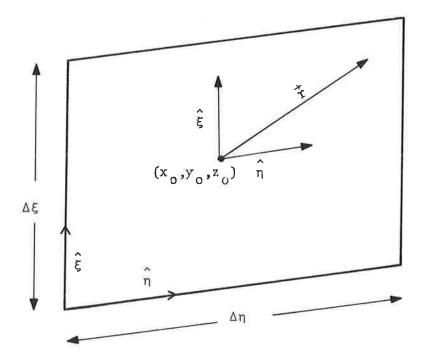


FIGURE 2. REPRESENTATION OF RECTANGULAR FACET

where

$$F_1 = n_2 \times (Y_1 - y) + [n_1 x + n_3 (z - h)] (X_1 - x)$$
 (27)

$$D_{1} = \left[x^{2} + (y-y_{a})^{2} + (z-h)^{2}\right]^{\frac{1}{2}}$$
 (28)

$$R_{1} = \left[(X_{1} - x)^{2} + (Y_{1} - y)^{2} + (Z_{1} - z)^{2} \right]^{\frac{1}{2}}$$
 (29)

Clearly, the distances D_1 and R_1 denote, respectively, the length of the vector \overline{D}_1 from the antenna to a point (x, y, z) on the surface of the facet and the length of the vector \overline{R}_1 from the point (x, y, z) to the receiver. From Figure 2, it is seen that the vector D₁ can be represented as follows:

$$\overrightarrow{D}_1 = \overrightarrow{D}_{10} + \overrightarrow{r} \tag{30}$$

where
$$D_{10} = X_0 \hat{e}_x + (y_0 - y_a) \hat{e}_y + (z_0 - h) \hat{e}_z$$
 (31)

and

$$\hat{\mathbf{r}} = \eta \hat{\eta} + \xi \hat{\xi} \tag{32}$$

and the ranges of the variables η and ξ are $-\frac{\Delta\eta}{2} \le \eta \le \frac{\Delta\eta}{2}$ and $-\frac{\Delta\xi}{2} \le \xi \le \frac{\Delta\xi}{2}$. From Equation (30), we have

$$D_{1} = D_{10} \left(1 + \frac{r^{2} + 2D_{10} \cdot r}{D_{10}^{2}} \right)^{1/2}$$
(33)

Normally, D_{10} is on the order of thousands of feet and r is of the order of tens of feet so that the radical in Equation (33) can be expanded in powers of r/D_{10} to give D_1 approximately as a function of η and ξ . To terms of second order in $r/\text{D}_{10}\text{,}$ we obtain

$$D_{1} \approx D_{10} + \eta \cos \alpha + \xi \cos \beta + \frac{r^{2}}{2D_{10}} - \frac{(D_{10} \cdot r)}{2D_{10}}$$
 (34)

where

$$\hat{D}_{10} = \vec{D}_{10}/D_{10} \tag{35}$$

$$\cos\alpha = D_{10} \cdot \hat{\eta} \tag{36}$$

$$\cos \beta = \hat{D}_{10} \cdot \hat{\xi} \tag{37}$$

The numbers $\cos\alpha$ and $\cos\beta$ are the direction cosines of D $_{10}$ relative to $\hat{\eta}$ and $\hat{\xi}$. In the Fraunhofer approximation which we will use, it is assumed that the quatratic terms of Equation (33) are small compared to a wavelength for any point (x, y, z) on the surface of the facet so that we can write to a good approximation:

$$D_1 = D_{10} + \eta \cos \alpha + \xi \cos \beta \tag{38}$$

It can be shown that the largest value attainable by the quadratic terms in Equation (33) on the facet's surface is

$$\Delta = \frac{\left(\Delta \eta\right)^2}{8 D_{10}} \sin^2 \alpha + \frac{\left(D\xi\right)^2}{8 D_{10}} \sin^2 \beta + \frac{\Delta \eta \Delta \xi}{4} \left| \cos \alpha \cos \beta \right| \tag{39}$$

Equation (38) should provide a good approximation for the distance D_1 as a function of η and ξ as long as $\Delta \eta$ and $\Delta \xi$ are chosen so that Δ in Equation (39) is small compared to the wavelength λ . We will return to this important requirement later.

Similarly, the vector \vec{R}_1 can be represented as follows:

$$\vec{R}_1 = \vec{R}_{10} - \vec{r} \tag{40}$$

where
$$\vec{R}_{10} = (X_1 - X_0) \hat{e}_x + (Y_1 - Y_0) \hat{e}_y + (Z_1 - Z_0) \hat{e}_z$$
 (41)

Thus, in the Fraunhofer approximation, we find

$$R_1 \simeq R_{10} - \eta \cos \alpha - \xi \cos \delta \tag{42}$$

$$\cos \alpha = {\stackrel{\wedge}{R}}_{10} {\stackrel{\wedge}{\eta}}$$
 (43)

$$\cos \delta = R_{10} \delta \tag{44}$$

where
$$\hat{R}_{10} = \hat{R}_{10}/R_{10}$$
 (45)

In Equation (42), we have ignored terms of second order and higher in r/R_{10} .

One final approximation which we will make is that $\Delta \, \text{n}$ and $\Delta \, \pmb{\xi}$ are sufficiently small compared with D_{10} and R_{10} , that the variation of the amplitude function $\text{F}_1/\text{D}_1^{\ 2}\text{R}_1^{\ 2}$ in Equation (26) can be ignored This is normally an excellent approximation. Consequently, we will assume that for any point on the surface of the facet we can write

$$F_1/D_1^2 R_1^2 \approx F_{10}/D_{10}^2 R_{10}^2$$
 (46)

where ${\rm F}_{10},~{\rm D}_{10},$ and ${\rm R}_{10},$ are the values of ${\rm F}_1,~{\rm D}_1,$ and ${\rm R}_1$ at the center point $({\rm x_o},~{\rm y_o},~{\rm z_o})$.

Substituting Equations (38), (42), and (46) into Equation (26) yields the following approximate expression for I_1 ;

$$I_{1} = \frac{F_{10}}{D_{10}^{2}R_{10}^{2}} e^{ik(D_{10}+R_{10})} \int_{-\Delta \eta/2}^{+\Delta \eta/2} e^{ik\eta(\cos\alpha - \cos\gamma)} d\eta \int_{e^{ik\xi}(\cos\beta - \cos\delta)} d\xi - \Delta \eta/2$$
(47)

Performing the integrals indicated in Equation (47), we finally obtain the following approximate expression for I_1 :

$$I_{1} = \frac{4F_{10}}{D_{10}^{2}R_{10}^{2}} \frac{e^{ik(D_{10}+R_{10})}}{k^{2}} \times \frac{\sin\left(\frac{k\Delta\eta}{2}(\cos\alpha-\cos\gamma)\right) \sin\left(k\frac{\Delta\xi}{2}(\cos\beta-\cos\delta)\right)}{(\cos\alpha-\cos\gamma)(\cos\beta-\cos\delta)}$$
(48)

A similar treatment of the integrals I_2 , I_3 , and I_4 ultimately lead to the following approximate expression for the scattered magnetic field intensity at the receiver due to the rectangular facet:

$$H_{SZ}(\vec{r}_{1}) = \frac{J_{o}}{2\pi^{2}} \frac{F_{10}e^{ik(D_{10}+R_{10})}}{D_{10}^{2}R_{10}^{2}} \frac{\sin(kA\Delta\eta/2)\sin(kB\Delta\xi/2)}{A B}$$

$$-\frac{J_{o}}{2\pi^{2}} \frac{F_{20}e^{ik(D_{20}+R_{10})}}{D_{20}^{2}R_{10}^{2}} \frac{\sin(kA_{1}\Delta\eta/2)\sin(kB_{1}\Delta\xi/2)}{A_{1}B_{1}}$$

$$-\frac{J_{o}}{2\pi^{2}} \frac{F_{10}e^{ik(D_{10}+R_{20})}}{D_{10}^{2}R_{20}^{2}} \frac{\sin(kA_{2}\Delta\eta/2)\sin(kB_{2}\Delta\xi/2)}{A_{2}B_{2}}$$

$$+\frac{J_{o}}{2\pi^{2}} \frac{F_{20}e^{ik(D_{20}+R_{20})}}{D_{20}^{2}R_{20}^{2}} \frac{\sin(kA_{3}\Delta\eta/2)\sin(kB_{3}\Delta\xi/2)}{A_{3}B_{3}}$$

$$(49)$$

where F_{10} , F_{20} , D_{10} , D_{20} , R_{10} and R_{20} denote the values of F_1 , F_2 D_1 , D_2 , R_1 , and R_2 at (x_0, y_0, z_0) and the parameters A, A_1 , A_2 , A_3 , B, B_1 , B_2 , and B_3 are defined as follows:

$$A = \cos\alpha - \cos\gamma \qquad B = \cos\beta - \cos\delta \qquad (50)$$

$$A_1 = \cos \alpha_1 = \cos \gamma \qquad B_1 = \cos \beta_1 = \cos \delta \tag{51}$$

$$A_2 = \cos\alpha - \cos\gamma_1 \qquad B_2 = \cos\beta - \cos\delta_1 \tag{52}$$

and the angles appearing in Equations (50) through (53) are defined as follows:

$$\vec{D}_{10} = x_0 e_x + (y_0 - y_a) e_y + (z_0 - h) e_z$$
 (54)

$$\frac{1}{R_{10}} = (X_1 - X_0) e_x + (Y_1 - Y_0) e_y + (Z_1 - Z_0) e_z$$
(55)

$$D_{20} = x_0^{\land} e_x + (y_0 - y_a) e_y + (z_0 + h) e_z$$
 (56)

$$\vec{R}_{20} = (X_1 - X_0) e_x + (Y_1 - Y_0) e_y - (Z_1 + Z_0) e_z$$
(57)

and

$$\cos \alpha = \frac{\overline{D}_{10}}{\overline{D}_{10}} \cdot \hat{\eta} , \cos \beta = \frac{\overline{D}_{10}}{\overline{D}_{10}} \cdot \frac{\hat{\xi}}{\xi}$$
 (58)

$$\cos \alpha_1 = \frac{\overline{D}_{20} \cdot \hat{\gamma}}{\overline{D}_{20}} \quad , \quad \cos \beta_1 = \frac{\overline{D}_{20} \cdot \hat{\xi}}{\overline{D}_{20}}$$
 (59)

$$\cos \alpha = \frac{\overline{R}_{10} \cdot \hat{\eta}}{R_{10}} , \quad \cos \delta = \frac{\overline{R}_{10} \cdot \hat{\xi}}{R_{10}}$$
(60)

and
$$\cos \alpha_1 = \frac{\overline{R}_{20} \cdot \hat{\gamma}}{R_{20}}$$
, $\cos \delta_1 = \frac{\overline{R}_{20} \cdot \hat{\xi}}{R_{20}}$ (61)

The total scattered field at the receiver is obtained by summing up the contributions from all of the rectangular facets which make up the surface of the scattering structure. Segmenting the surface of a scatterer into rectangular facets and then applying the approximate closed form solution in Equation (49) to each facet represents a far more efficient procedure for calculating the scattered field at the receiver than attempting to directly numerically integrate Equations (14) through (18).

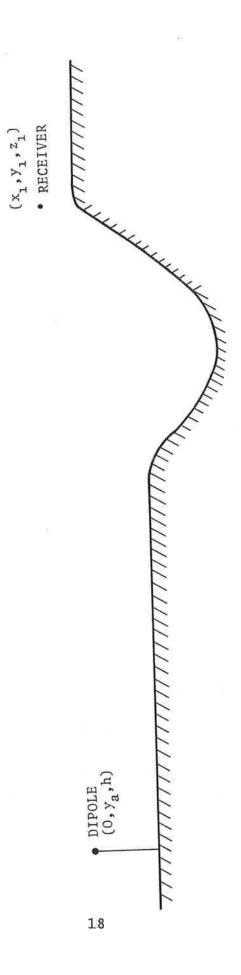
We now return to the question raised earlier of the size restrictions upon the facets. Clearly, one would like to maximize the area of each facet in order to minimize computer running time. The principal restriction upon the size of the facets is that Δ_{η} and $\Delta\xi$ must not be so large that the Fraunhofer approximation, upon which the approximate formula (49) is based, is violated. That is, the quadratic path length difference terms like those in Equation (34)

must be small compared to the wavelength λ for all points on the surface of the facet to which Equation (49) is to be applied. Thus, in choosing appropriate values for $\Delta\xi$ and $\Delta\xi$, we are essentially faced with an optimization problem with a constraint. Specifically, we want to maximize the area $\Delta\eta\Delta\xi$ of the facet while keeping the quadratic path length difference terms less than some prescribed value. For example, for the path length D₁, the maximum value of the quadratic path length difference term is given by

$$\Delta = \frac{(\Delta \eta)^2}{8 D_{10}} \sin^2 \alpha + \frac{(\Delta \xi)^2}{8 D_{10}} + \frac{\Delta \eta \Delta \xi}{4} \left| \cos \alpha \cos \beta \right|$$
 (62)

Depending upon the sign of $\cos\alpha\cos\beta$, this value will be achieved at two of the four corners of the rectangular facet. At all other points, the quadratic path length terms will be less than Δ , but always positive. We can optimize our choice of $\Delta\eta$ and $\Delta\xi$ by maximizing $\Delta\eta$ $\Delta\xi$ subject to the condition that Δ in Equation (62) be some fractional part of a wavelength. This problem has been solved in closed form using Lagrange multipliers. The optimization procedure must be applied to the quadratic path length difference terms affecting all four distances (D₁, D₂, R₁, R₂) and then from the resulting set of solutions the smallest values of $\Delta\eta$ and $\Delta\xi$ are chosen. A subroutine has been written to carry out the optimization procedure.

Of course, the scattered fields must be calculated for each dipole in the array, the parameter \mathbf{J}_0 being varied to take into account the exact phase and amplitude ratios among the various elements and, along with \mathbf{J}_0 , the height h and y-offset \mathbf{y}_a of each element. Our program automatically offsets the elements of an array to correct for the effects of proximity phase lag. The y-coordinates of the elements are adjusted so that all dipoles are at the same slant distance from the touchdown point on the runway directly opposite the array.



TYPICAL PROFILE OF ONE-DIMENSIONAL TERRAIN VARIATION FIGURE 3.

will generate the ground reflected signals at the carrier frequency. 90Hz modulated frequency, and the 150Hz modulated frequency. These signals are then combined with the incident signals directly from the antennas to give what we have been calling the direct fields, the fields which would be measured at the receiver assuming that no scatterers (hangars, hillsides, etc.) were present. These direct signals are then combined with the scattered signals calculated using the techniques of Part A and the resulting total signals are used in Equation (65) to calculate C.D.I. Again, if the terrain profile is perfectly flat or nearly so, Equation (63) can be used to calculate the direct signals.

3. NUMERICAL RESULTS

Before presenting some sample numerical results illustrating the predictions of our model, we will first review the characteristics of the three main types of image glide slope arrays; the null reference, the sideband reference, and the capture effect arrays.

The null reference array is the simplest of the image glide slope systems. It consists of two transmitting antennas whose heights are in the ratio of 2:1. The upper antenna is fed sideband only signal, the 150 Hz and the 90 Hz components being of equal amplitude but 180° out of phase. The lower antenna is fed both carrier and sideband. The carrier signal is nominally 40% modulated. The ratio of the sideband currents in the upper antenna to the sideband currents in the lower antenna is typically 0.3. The sideband ratio of 0.3 would nominally yield a 1.4° course width (full deflection 0.7° above or below the glide path). Let

 $^{\rm I}$ _c, $^{\rm I}$ ₁₅₀, and $^{\rm I}$ ₉₀ denote the current amplitudes fed to the antennas at the carrier frequency, the 150 Hz modulated frequency, and the 90 Hz modulated frequency, respectively. The null reference array parameters discussed above can be summarized as follows:

Carrier Antenna	Sideband Antenna
Height = h	Height = 2h
$I_c = 1$	$I_c = 0$
$I_{150} = 0.4$	$I_{150} = 0.12$
$I_{90} = 0.4$	I ₉₀ =-0.12

Note that all current amplitudes have been normalized relative to the carrier current amplitude (I_c = 1).

The sideband reference array employs two transmitting dipoles whose heights are in the ratio of 3:1. If the lower antenna is positioned at 1/2 the height of the lower antenna of the null reference array and if the upper antenna is positioned at 3/4 of the height of the upper antenna of the null reference array, the same glide angle is produced. Modulated carrier (40% modulated) is fed to the lower antenna. Both antennas are fed separate sideband signals. The separate sideband signals fed to the two antennas are equal in amplitude but are 180° out of phase. The amplitude ratio of the separate sideband signal to the carrier sideband signal is typically 0.3. This ratio produces a nominal course width of 1.4° as in the case of the null reference array. The sideband reference parameters are summarized below.

Lower Antenna	Upper Antenna
Height = h	Height = 3h
$I_c = 1$	$I_{c} = 0$
$I_{150} = 0.28$	$I_{150} = 0.12$
$I_{90} = 0.52$	$I_{90} = -0.12$

Note that the I_{150} and I_{90} current amplitudes given above for the lower antenna represent the sums of the carrier sideband and separate sideband signsls (I_{150} = .4 - .12 = .28, I_{90} = .4 + .12 = .52). All current amplitudes have been normalized relative to the carrier signal amplitude.

The capture effect glide slope array consists of three transmitting antennas whose heights are in the ratios of 1:2:3. If the lower and middle antennas are set at the same heights as the null reference antennas, the same glide angle is produced. We will not treat the clearance signal which provides a strong fly up signal at low approach angles but has little effect upon the glide angle and course width. Concerning the primary signal, the modulated carrier is fed to both the lower and middle antennas. The modulated carrier fed to the middle antenna has half the amplitude and is 180° out of phase with the modulated carrier fed to the lower antenna. carrier signals are nominally 40% modulated. In addition, all three antennas are fed separate sideband signals. The separate sideband signals fed to the lower and upper antennas have half the amplitude and are 180° out of phase with the separate sideband signal fed to the middle antenna. The ratio of the separate sideband signal fed to the middle antenna to the carrier sideband signal in the lower antenna is typically 0.3. This ratio yields nominally a course width of 1.4° as in the case of the null reference array. capture effect array parameters are summarized below.

Lower Antenna	Middle Antenna
Height = h	Height = 2h
$I_c = 1$	$I_{C} = -0.5$
$I_{150} = 0.34$	$I_{150} = -0.08$
$I_{90} = 0.46$	$I_{90} = -0.32$

Upper Antenna

Height = 3h

$$I_c = 0$$

 $I_{150} = -0.06$
 $I_{90} = 0.06$

Note that the $\rm I_{150}$ and $\rm I_{90}$ current amplitudes given above represent the sums of the carrier sideband and separate sideband signals. Again, all values have been normalized relative to the carrier amplitude in the lower antenna.

It should be noted here that our computer program automatically offsets the elements of each glide slope array to correct for the effects of proximity phase lag. The y coordinates of the elements are adjusted so that all dipoles are at the same slant distance from the touchdown point on the runway directly opposite the array. For each array, one element is held fixed while the other elements are offset relative to the fixed one. For the null reference and sideband reference arrays, the lower element is held fixed while for the capture array, the middle element is held fixed. The amount of the offset for each dipole can be calculated approximately in the following manner. Let y_a and h denote, respectively, the y-coordinate and height of the fixed antenna element. The y-displacement, ε , of any other element in the array relative to the fixed one is given approximately by:

$$\varepsilon = \frac{h^2 - H^2}{2y_a} \tag{66}$$

where H is the height of the element which is to be offset.

For the present study, the wavelength λ was set at 3 feet. All arrays were positioned 300 feet from the centerline of the runway. The heights of the array elements were set at 14.33 feet and 28.66 feet (approximately 5λ and 10λ) for the null reference array, 7.17 feet and 21.5 feet (approximately 2.5λ and 7.5λ) for the sideband reference array, and 14.33 feet, 28.66 feet, and 42.99 feet (approximately 5λ , 10λ , and 15λ) for the capture effect array. Under normal circumstances, a glide angle of 3° would be produced by all three arrays.

To illustrate the performance of our multipath model, the glide slope course deragation caused by a large, flat reflecting surface (perhaps the side of a large hangar or the side of a hill) has been calculated. The structure was taken to be 300 feet long and 100 feet high and was oriented parallel to the runway centerline. The structure was positioned 200 feet from the runway centerline in the y-direction, on the opoosite side of the centerline from the glide slope array. The near edge of the structure was displaced 1000 feet in the x-direction from the touchdown point. The geometry of the configuration is depicted in Figure 4.

In Figures 5, 6, and 7 are plotted the C.D.I. signal as a function of the horizontal distance from touchdown for each of the three arrays for the arrangement depicted in Figure 4. In each case, the receiver followed the nominal 30 zero C.D.I. hyperbola. The results for all three arrays are very similar. Our model predicts large high frequency, alternating fly up and fly down signals near the end of the flight path roughly between 1700 feet and 1800 feet from touchdown. This region in fact corresponds almost exactly to the region of specular reflection as predicted by geometrical optics. Note that the excursions are large enough to saturate the receiver over much of the affected range (full scale deflection = 150 microamps). The excursions are clearly highly localized and damp out very rapidly as the aircraft leaves the specular reflection zone. The narrowness of the pattern (i.e., the lack of broadening due to diffraction and the rapid damping of the side lobes) is undoubtedly attributable to the large size of the scatterer relative to the scattering wavelength.

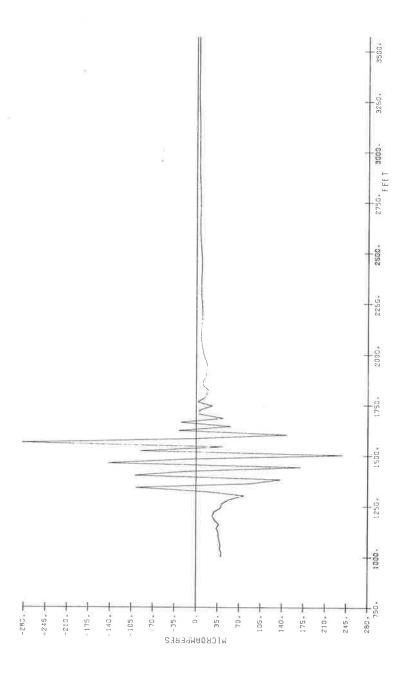
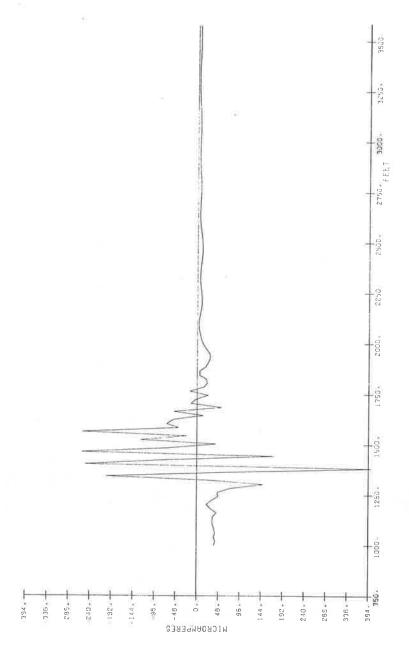


FIGURE 6. SIDEBAND REFERENCE ARRAY, FLYABILITY RUN



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	C THIS SECTION INPUTS THE GROUND STRIPS DECSRIPTIONS
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	READ(20,1000) ILABL
	WRITE(1,1000) JLABL
	READ(20) K,X1,Z1,X2,Z2
	CALL RELEAS (20)
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	2001 FORMAT(" INPUT ANTENNA FILE NAME:", \$)
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	2000 FORMAT(A5)
	CALL IFILE(20,1LBL)
	REAU(20,1000) ILABL
	WRITE(1,1000) ILABL

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COMHON /REC/ RX(4), MSIZE
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EUD NSIZE /RFC **4218N** 000000000001 0+/ /REC SUBPROGRAMS INPUT 36 CONSTANTS ARRAYS SCALARS COMMON JUVWI RX RX С INPUT JOVWI NSIZE REC RX

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XI(1) THE X-COORDINATE OF THE LEADING FUCE UP THE LTF STRIP
ZI(1) THE Z-COORDINATE OF THE LEADING FUCE UP THE LTF STRIP
XZ(1) THE X-COORDINATE OF THE ENDING EDGE OF THE LTH STRIP
ZZ(1) THE Z-COORDINATE OF THE ENDING EDGE OF THE LTH STRIP
SUBMENUTINE SCAT
                                                                                                                                                                                                                                                           THESE ARE INITIAL VALUES FOR THE PARAMETERS USED IN SHADOWING
                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                    IF THE PECIEVER IS LOCATED OWER THE MIDDLE PORTION OF A STRIP
THE STRIP WILL BE INTEGRATE ONLY UP TO THE VALUE OF THE
PECEIVEP X-COORDINATE
IF(XXZ .LE. RX) GO TO S
ZZZ=ZXI+(RX-XXI)*(ZZZ-ZZI)/(XXZ-XXI)
                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                    THIS SECTION DUES THE SHADOWING. IF PART OR ALL OF THE STRIP IS IN THE SHADOW OF A PREVIOUS STRIP, THIS STRIP WILL BE ELIMINATED OR MASKED TO GIVE THE EFFECT OF SHADOWING.
                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                    STRIPS HAS REACHED THE RECEIVER LUCATION, IF IT HAS THE SLOWEATION IS STUDPED, THIS IS TO GIVE THE REFECT OF FOWARD LUDNING RECEIVER ANTENN.
               THIS SURPOUTTRE SUNS THE EFFECTS OF THE STRIPS THAT PARE UP HE GHOUNDSUPFACE. THERE APE "N" STRIPS DESCRIBED IN COP 1079
                                                                                                                                                                                                     COMMON /GROUND/ P.XI(20),ZI(20),XZ(0/20),ZZ(0/20),ILL
COMMON /SEG/ XX1,ZZ1,XX2,ZZ2,4
                                                                                                                                                                                                                                                                                                                                                                                        THESE ARE THE VALUES TO BE USED IN THE STRIP INTEGRATION SURROUTINE
                                                                                                                                                                                                                                                                                                                              THIS IS THE LOOP OVER THE STRIPS DO 1 I=1,K
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PHIE=(AZ-ZZ2)/DELX
IF(SLOPE .LT. 0.) GO TO 3
                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                   IFC XXI .GF. RX) GO TO 6
                                                                                                                                                               COMMON /REC/ PX, PY, PZ
                                                     GRUHND AS FULLUMS:
                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                             DELX=XX2-AX
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221=21(1)
221=22
222=22(1)
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CUNTIAUE
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				. A1 X2 XX2		774# * 3 "]	
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		35 15 (1)		AX 2.1 2.2.1		1 22.2 PHI AY	22
		TO HP AND HI		/PEC /+2 /Gp0U4D/+1 /SEG /+0		1	į
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		ATE NVEP 1F WILL		3 X X 1 X X X X X X X X X X X X X X X X		PHI K X X Z X Z X Z H Z	x2
IF (PHIE .GF. SLOPE) GU TU 1 PHIE=(AZ-ZZ1)/(XX1-AX) IF (PHIB .LE. SLOPE) GU TU 3 IF (XX1 .E.Q XX2) GU TU 4 A=(ZZ2-ZZ1)/(XX2-XX1)	17)+AX	THIS SUBRUUTIVE WILL INTEGRATE OVER THE SIRIP THE LEX "GAIN" EFFECT OF THIS STRIF WILL BE ADDED TO HP CALL SUM CALL SUM COMPINUE CONTINUE FRITHN FRITHN FRITHN		/REC /+1 /GROUTAD/+0 /GROUTAD/+123 /SEG /++		*	25
IF (PHIE , GE, SLOPE) GU T PHIE=(AZ-ZZ1)/(XX1-AX) IF (PHIB , LE, SLOPE) GN IF (XX1 , EQ, XX2) GU TN A=(ZZ)-ZZ1,/(XX2-XX1)	R=ZZ1=A+(XX1=AX) XX1=(AZ=B)/(A+SUNPE)+AX ZZ1==SUOPE+(XX1=AX)+AZ	SUBRUUTIME IN EFFECT VALL *		z z z Z		SLOPE 221 AX RY	21
IF (PHIE PHIE IF (XXI	R=ZZ1-A+ XX1=(AZ- ZZ1=-SL	THIS SUBRUI FDEX "GAIN" EN COMMON "VALL" CALL SUM SLOPE=PHIE CONTINUE FRIEN	5 201400000000	/REC /+0 /ANT /+2 /GRUDND/+7b /SEG /+3			
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THIS SUBBOUTINE INTEGRATES HIVER THE SUBFACE STRIP DEFINED
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COMMON "REC", THE VARIABLES ANE AS FOLLOWS:

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                                                                                                                                                                                                                                                                                                                                                                             LEADING EDGE OF STRIP'S X-CUMBBLHATE
LEADING Z-COMBINATE
                                                                                                                                                                                                                                                                                                     RECEIVER Z-CUMPUTMATE
PEL PART UF 'GAIM' FACTOR
IMAGIMARY PART OF 'GAIM' FACTOR
                                                                                                                                                                                                                                                                                                                                                                                                                             TRAILING EDGE X-COORDINATE
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                                                                                                                                                                                                    TWU-PI/LAMBDA
TWO-PJ (DUUBLE PRECISION)
                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                SUMPOUTINE SUM
COMMON /SEG/ X1,21,X2,22,h
DOUNLE PRECISION A1,42,81,82,X1,AY2
PEAL LAMBDA
REAL LAMBDA
                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                   DOUBLE PRECISION AK.DPI.DR
COMMON ANTAKAYAY.
COMMON ARECRE, MY. PZ.
COMMON AVALARE, I
                                                                                                                                                                                                                                                     RECEIVER X-COORDINATE PECFIVER Y-COORDINATE
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Sh=Z2-Z1
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C ESNACIOR) TEMP DESCRARA I SORVEDE AND TEMP TEMPORALIS IS THE PHASEE ANGLE "MODULO IND PJ- E THIS IS THE PHASEE ANGLE "MODULO IND PJ- DECA SEANT(I, A+1, A+1, A+1, A+1) C) C THE SEANT (I, A+1, A+1, A+1) C) C THES ARE THE REAL AN IMAGINARY PARTS OF THE THE SUMMATION. OF THE SEANT (I, A+1, A+1) C) C THESP ARE THE REAL AN IMAGINARY PARTS OF THE THE SUMMATION. OF THE SEANT (I, A+1) C) C OF THE APPROXIMATION TO THE INTEGRAL JH-10(1-JAH1) TEMP THE SEANT (I, A+1) C OF THE PHASE FUNCTION OF THE SEANT (I, A+1) C OF THE PHASE FUNCTION F PARSICAPABBLY.	22.	DR=DSQRIC	RLE(TEMP) *DBLE(TEMP) + DBLE(AY*AY))			
THIS IS THE PHASEE ANGLE "MODULU IND PI- D=0+-Chalceloat(I))**PPI D=0+-Chalceloat(I))**PPI D=0+-Chalceloat(I))**PPI S=0+-Chalceloat(I))**PPI S=0+-Chalceloat(I) S=0+-	:3	C=SNGI, CDR	/TE4P			
THIS IS THE PHASEE ANGLE "MODULO IND PI- F=DH-PUBLE(FLOAT(I))*PPI B=C.** B=C.** S=SAPT((1./A+1./R)/C)/C G=A1/(10+05*R*R\$) C THIS IS THE PRADE PAT UP THE INTEGRAND FUN THE THE SURANTIUM VARIABLE VALUE OF XL ANH=G*CF-TEMP*SF C THESE ARE THE REAL AN INAGINARY PARTS OF THE THE SURANTIUM. THIS IS THE INAGINARY PART THESE ARE THE REAL AN INAGINARY PARTS OF THE INTEGRAN. C THE IMAGINARY PERPONE C THE IMAGINARY PERPONE C THE IMAGINARY PART THIS IS THE INAGINARY PART THIS IS THE BEAL AN INAGINARY PARTS C OF THE IMAGINARY PART THIS IS THE BEAL AN INAGINARY PARTS C OF THE IMAGINARY PART THIS IS THE DERIVATIVE OF THE PHASE FUNCTION ODI-JII - ODI-JIE C FP IS THE DERIVATIVE OF THE PHASE FUNCTION FP =ANS((AP\$BP)/C)	24	DR=DR*BK		39 II St. 50 ST. 10 ST.		
THIS IS THE PHASEE ANGLE "MODULU TWO PJ" PECHAL SESOPT((1,/A+1,/H)/C)/C GASIOT((2AK#8)) THIS IS THE AMPLITUDE FUNCTION FERDEC(CAK#8) C THIS IS THE PRAL PART OF THE INTEGRAND FOR THE SF=SIN(F) SF=SIN(F) C THIS IS THE PRAL PART OF THE INTEGRAND FOR THE JNIGGEF-FRAMSEE THIS IS THE PRAL AN INAGINARY PARTS OF THE THE SUBMATIO. ANGESF-FERMSEE THES ARE THE REAL AN INAGINARY PARTS OF THE INTEGRAL THESE ARE THE REAL AN INAGINARY PARTS OF THE INTEGRAL OF THE THAPTOLOS WINTIG IN THE APPROXIMATIOUT THE INTEGRAL JNIGGEF-FRAMSEE C THESE ARE THE PRAL AN INAGINARY PARTS OF THE INTEGRAL OF THE THAPTOLOS WINTIG IN THE APPROXIMATIOUT THE INTEGRAL OF THE THAPTOLOS WINTIG IN THE APPROXIMATIOUT THE INTEGRAL OF THE THAPTOLOS WINTIG IN THE APPROXIMATIOUT THE INTEGRAL OF THE THAPTOLOS WINTIG IN THE PHASE FUNCTION OF THE THAPTOLOS WINTIG IN THE PHASE FUNCTION OF THE THAPTOLOS WINTIG OF THE PHASE FUNCTION FPHASE(ARASED/G	52					Į,
THIS IS THE PHABER ANGLE "MODULU TWO PITTOR-" FECH-CHECFLOAT(I) **PPI C THIS IS THE PART OF THE INTEGRAND FOR THE C THIS IS THE PEAL PART OF THE INTEGRAND FOR THE C THIS IS THE PEAL PART OF THE INTEGRAND FOR THE C THIS IS THE PEAL PART OF THE INTEGRAND FOR THE C THIS IS THE PEAL WALLE OF XL JNHEGECF-PREPARE C THIS IS THE PEAL AN INAGINARY PART C THIS IS THE INAGINARY PART C THE INADIAL PART OF THE PRADALATION TO THE INTEGRAL JI = J1+(J01+JNI)**FEMP JI = J1+(J01+JNI)**FEMP JI = J1+(J01+JNI)**FEMP S JOI = JNI RP = LANCE ARE SESSED'A RP = KALANCE ARESESED'A FP IS THE DERIVATIVE OF THE PHASE FUNCTION FP = ANSI (APABP)/C)	56					
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READ(20,10n0) ILABL

WHITE(1,100) ILABL

FORMATCAS,F)

READ(20) LAMBDA,NEL,(X(I),Y(I),Z(I),CFI(I),CFZ(I),CF3(I),I=1,FEL)

WHITE(1) LAMBDA,NEL,(X(I),Y(I),Z(I),CFI(I),CFZ(I),CF3(I),I=1,FEL)

DAK=DPI/DRIE(LAMBDA)
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PURMATIC OUTPUT FILE NAME: '. S)
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t, 4 DOUGHE PRECISION FUNCTION DIST(X,Y)
DUMHE PRECISION R,TEMP
R=0.
DO 1 1=1,3
TEMP=X(1)-Y(1)
R=R+TEMP*TEMP
DIST=NSORT(R)
RETURN 24 29 57 rr 4 4 DFM.2 DFAM.2 DSORF GLOBAL DUWMIES SUBPROGRAMS 56 DIST 60 56 SCALARS ARRAYS 'n × 1 P DIST DSURT I R TEMP

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TEMP

20 	/XA,TA,ZA,LAMBDA,DAK,DPI (P11,P(1,1)),(P12,P(1,2)),(P21,P(2,1)) (P24,P(2,4)),(P31,P(3,1)),(P34,P(3,4)) TRUE,/ LL NORMAL(P11,P12,P(1,4),P(1),TEMP)		
1 SUBROUTINE INTRZ(CTEMP) 2 DATA DX,DY,23,40,4 3 DIMENSION XTA(3),274(3), 5 COMPLEX A,B 6 COMPLEX A,B 7 COMPLEX A,B 7 COMPLEX CTEMP,CCT,CCX,CCX,CCX,CCX,CCX,CCX,CCX,CCX,CCX	13 COMMON ANT/XA,TA,ZA,LAMBDA,DAK,DP1 14 EQUIVALENCE (P11.P(1.1)) (P12.P(1.2)), (P21.P(2.1)) 15 EQUIVALENCE (P24.P(2.4)), (P31.P(3.1)), (P34.P(3.4)) 16 LOGICAL TEST 17 DATA TEST,TRUE, 18 F(TEST) CALL NURMAL(P11.P12.P(1,4),B(1),TEMP) 20 ZAZZ,RASE. 21 ZAZZ,RASE.		49 DG1Z=ZTA(3)*DELZ 50 DG 1 IXX=1,1X 51 FX=FLIX)*S 52 XS=P(1,1)*FX*DG0X YS=P(2,1)*FX*DG0X

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P2=D20
F10=N2&XSe(YR-XS)+(N1*XS+K3*(ZS-ZA))*(XR-XS)
F20=F10+N3+(XR-XS)*ZAZ
COSA=(XTA(1)*CC7+XTA(2)*(YS-YA)+XTA(3)*(ZS-ZA))/D1
COSA=(XTA(1)*CC7+XTA(2)*(YS-YA)+XTA(3)*(ZS-ZA))/D2
COSA=(XTA(1)*CC7+XTA(2)*(YR-YS)+XTA(3)*(ZS-ZS))/R2
COSG=(XTA(1)*CC6+XTA(2)*(YR-YS)+XTA(3)*(ZS+ZS))/R2
COSB=(ZTA(1)*CC7+ZTA(2)*(YS-YA)+ZTA(3)*(ZS+ZS))/R2
COSB=(ZTA(1)*CC7+ZTA(2)*(YS-XA)+ZTA(3)*(ZS+ZA))/D1
COSB=(ZTA(1)*CC6+ZTA(2)*(YS-XA)+ZTA(3)*(ZS+ZA))/D2
COSD=(ZTA(1)*CC6+ZTA(2)*(YS-YA)+ZTA(3)*(ZS+ZA))/D2
COSD=(ZTA(1)*CC6+ZTA(2)*(YR-YS)+ZTA(3)*(ZS+ZA))/R2
2.5=p(3,1)*px*bC02

0.0 2 122=1,12

F2=F1.04T(L22)-.5

X5=X4=F2+0G1X

Y5=X5+F2+0G1X

Y5=X5+F2+0G1X

CC5 XR=XA

CC5 XR=XA

TEMP=ZR-ZS

CC7 XS-XA

TEMP=ZR-ZS

TEMP=ZS-TS

R12-DR1

R2-DR2

R12-DR3

R12-DR1

R2-DR3

R12-DR3

R12-DR3

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CH2=COSA1-CUSG
CH3=COSA1-COSG1
C=CUSH-CUSB
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D10=D10+DAK
D20=D20+DAK
R10=R10+DAK
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TEMPET10*SIN(AX*CH*DELX*,5)*SIN(AX*C*DELX*,5)/(CH*C*D12*F12)

TEMPET10*SIN(AX*CH*DELX*,5)*SIN(AX*CT*DFLX*,5)/(CH*CT*D12*F22)

CTEMPECTEMP-TEMPECEXP(CMPLX(0,,D1H2))

CTEMPECTEMP-TEMPECEXP(CMPLX(0,,D1H2))

TEMPET20*SIN(AX*CH*DELX*,5)*SIN(AX*CZ*DFLX*,5)/(CH3*CZ*DZZ*F12)

TEMPET20*SIN(AX*CH3*DELX*,5)*SIN(AX*CZ*DFLX*,5)/(CH3*CZ*DZZ*F12)

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200 ENGUDE (9,2006,1LABL(1)) ILBU 2006 FURRATES, PIDENTS,                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                    RXFT
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                                     C THIS SECTION SIMULATE THE REFECT OF THE FLECTICAL AND C MECHANICAL TIMERIA* OF THE ILS C RECEIVER SYSTEM FOR DYNAMIC SIMULATION TOERT
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IF(ADDI GT. ADIN) ADIN=ADDI
FF(ADDI L. ADIN) ADN=ADDI
FF(ADDI .T. ADNN) ADN=ADDN
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	FORSE.	JAFF PEFINE	ALPHO.	ALPHI.	TFTLE EXIT	OFILE	FINNE.	FLUUT.	FILIST.	EMB.	PFAL	5.0.5	E > F	Cr2	BFITAS
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	ADDR	612		10	613		XICA	32		ADT.	3.1		AINTE	3.3	
	ADRX	36		ADPR	35		ADRE	3.7		PYFT	147		1111	-	
	RZFT	13		RTET	17		AIFT	23		AHFI	2.1		1. 4.	-	
-	PXMX	2		PYMN	2		RYMX	q		24	11		у-2н	71	
	RTMN	15		PT#X	16		AIMX	25		Alse	21		ARTIK	47	
	ARMN	25		PXI,T	4		RYLT	10		RZLI	14		RILT.	25	
	AILT	24		ARLT	30		CON	614		ADIL	3.4		ATTHE.	10	
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	OF ILE PLXT REAL REC	ELEAS	RIFI	THE	THE	RX	RXFT	XIL	RXMX	RY	YET	RYLT	YMX	2	7.42	ZLL	NW2	0	AU	VAL		2	, k				-	į	D 0	200P	201P	1000P	2000P	2001P	2002P	2003P	2006P

<i>7.</i> 8	, , , , , , , , , , , , , , , , , , , ,					
DATA ITYPE(3,2) DATA ITYPE, STATT, C VAL., UES., UTRANT, TC VA., LUES.) DATA FID. PRD. (CDID., THEUD.) DIMENSION SPACE(4), 1X(2,3) DATA SPACE/1.2.2.5.5. DATA BAX. PRGERE, 'ES., FEET., 'SECUN', 'DS') DATA PRX. PRY, PRY, PRT, PAI, DAP. DIMENSION 17DAT(33)	COMMON /PIXXN/NRP, PXMN, PXMX, PXFT, PXLT, PYMN, RYMX, RYFT, FYLT, EXMI, PZMX, PZFT, HZLT, RTMM, RTMX, PFFT, RTLT, FTLT, EXMIN, ALMX, ALFT, ARBIT, ARMY, ALMX, ALMY, ARMY, ARMY, ARBIT, ARMY, AND MANN, ARMY, ARMY, ARBIT, BCUJVALENCE (IPTDAT(1), MRP) DIMENSION ILARL(R) DIMENSION DIT(2000)	DIMENSION DX(2000), DY(2000) NAMELISI /FRED, YLENG, YDEL, YSC, DMIN, DMAK, DEL, IP, KSC CALL PLOTS(INIP, 360, 16) WHITE(5, 1006) FORMAI(* INPUT FILE NAME AND AXIS TYPES: ', s) HEAD(5, 1005) NAME, ISX, ISY, BUUND	FORMAT(AS, 1.1.F) IF(ISY, LT, 1) GD TD 204 IF(ISY, LT, 1) GD TD 204 IF(ISX, LT, 1) GD TD 204 IF(ISX, LT, 1) GD TD 204 IF(ISX, LT, 1) GD TD 204 GALU, PLOT(0,*12,*-3)	I=0 CALL IFILE(20, MAME) RRILE(3,1002) NRP RRILE(3,1002) NRP FORWATC THERE ARE', IS, 'RECEIVER PUINTS,',') FORWATC(14x,'MIN',9X,'MAX',9X,'FIRST',9X,'LAST',)	FORMAT(X,AS,1X,4F12,4) WPITE(3,1004) PRX,RXME,RXX,RXFT,RXLT RE(3,1004) PRX,RYMH,RYX,RYFT,RYLT RETE(3,1004) PRX,RYMH,RXX,RZFT,RZLT WRITE(3,1004) PRT,RTM,RTXX,RZFT,RZLT WRITE(3,1004) PRT,RTM,RTXX,RTFT,RTT RETE(3,1004) PAI,A1MN,A1MX,A1FT,A1LT WRITE(3,1004) PAR,AHMN,AHMX,ARFT,ARLT	WRITE(3,1004) PID,ADIN,ADIR,ADIL GASD=0. GASD=
-	H N M	1006	1005	A MERCHANIC	1004	101
- ~ m 4 N & ^ # o	12 13 14 15 17	118 20 21 22 23	24 27 29 30 30	3321	338 441 443 443	4 4 4 4 4 4 4 4 4 4 4 4 4 4 4 4 4 4 4

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TEMPESORT(X*X+Y*Y)

IF((TEMP .LT. 3500.) .OR. (TEMP .GT. 20720.)) Gn TU 60

GASU=GASU+C
GASU=GASU+C
GASU=GASU+C
GASU=GASU+CP
GACU=GASN+CP
GACU=GASN+1.
            Call SYMBOL(0.,0.,12,1LAML,90.,40)
Call Pint(3,0.,-3)
Call SYMBOL(0.,0.,12,1TYPE(1,1SY),90.,15)
Call Duf(2.,0.,-3)
REAU(20,1000,END=2) X,Y,Z,I,C,H,CD,PP
FORMAT(RF)
                                                                                                                                                                                                                                                                                                                                                                                                                                                               DNAX=DX(I)
DNAX=DX(I)
DNAX=AANI(DNAX,DX(I))
DNAX=AAAX((DNAX,DX(I))
I FURMAT(SX,3F)
IE ( 1 .LT. 2000) GU TO 1
IF ( 1 .LT. 2) GO TO 3
YLENG=MAXI(AIMX,ARS(BOUND), AIMN,8.)
IF (ARS(BOUND) .LT. 1,E-4) GO TN 10
                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                              YLENG=AMAXIIYLENG,ARMX,-ARMN)
YDEL=FLOAT(IFIX(YLENG/YLEN))
YLENG=YDEL&YLEN
YLENG=XIS3(0,0,-YLENG,YLENG,YDFL,YLEN,
I***ICROAMPERES,12,0,0,***YSC)
I****ICROAMPERES,12,0,0,***YSC)
                                                                                                                                                                                                                                                                                                 DY1(T)=RD
CUNTINUE
GO TO (200, 201, 202) ISX
DX(L)=ATM2(Z, SCRT(X*X+Y*X))*57,2953
GO TO 199
GU TO 199
DX(L)=T
GO TO 199
GO TO 199
                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                     PUM=10.**IP.

DO 120 J=1,4

DEL=SPACE(J)*POM
IT=IFIX(DMAX/DEL-1,)+1-IFIX(DMIN/DEL)
IF(IT, bT, ITIC) GO TO 121

CONTINUE
                                                                                                                                                                                                                                                                                                                                                                                                                                      IF(I .NE. 1) GO TO 198
DMIN=DX(I)
                                                                                                                                                                                                            GO TO (300,301) ISY
CONTINUE
FURMAT (8A5)
                                                                          1000
                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                             1001
100
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CALL SYMHOL(XXC,YYC,,12,RCD,90,*w2,LCH) IFFPWR,EQ.O.) HETIRN CALL WHERE (XX,YQ,XXZ,12,5H * 10,90,*w2,5) CALL WHERE (XX,YQ,XZ,12,5H * 10,90,*w2,5) X = XQ + (XXC, RR,XQ)*w2, X = YQ + (XYC,0R,YQ)*w1 CALL NUMBER(X,Y,09,PWR,90,*w2,-1) RETURN N=NCR/5,1000) AWAX,AMIN,DELA,PWR,(RCD(1),1=1,1) FRIURN PWRIF(S,1000) AWAX,AMIN,DELA,PWR,(RCD(1),1=1,1) Y,1X,4G,/,1X,13A5) CALL EXIT CALL EXIT CALL EXIT CALL EXIT		20 3	A BCD	rF1X		C	N W	IMORE	You	
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MEASUREMENT OF THE ATCRBS SURFACE INTERROGATION ENVIRONMENTS AT CHICAGO O'HARE AND LOS ANGELES INTERNATIONAL AIRPORTS

M. J. Moroney H. J. Glynn

U.S. Department of Transportation Transportation Systems Center Kendall Square Cambridge MA 02142



JULY 1976
INTERIM REPORT

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Systems Research and Development Service
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16. Abstract

The Transportation Systems Center is conducting a program to develop a surface surveillance sensor that uses replies from ATCRBS transponders. The operation of this system can be affected by surface interrogations at major airports where such a system might eventually be deployed. Consequently, tests were conducted at Chicago O'Hare and Los Angeles International Airports to measure the surface interrogation environment and to determine the number of interrogators causing surface transponders to reply.

This report describes the tests that were performed, presents the analysis of collected data, and offers conclusions pertinent to future operational ASTC systems.

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PREFACE

The test series described in this report is part of a comprehensive program at the Transportation Systems Center (TSC) to investigate and evaluate factors bearing on the development of a surface surveillance system that operates with replies from ATCRBS transponders. This program is being conducted by the Airport Surface Traffic Control (ASTC) Program Office at TSC and is sponsored by the Federal Aviation Administration (FAA) through the Systems Research and Development Service. One of the factors being investigated is the ATCRBS interrogation rate for vehicles on the surface of airports. The surface surveillance system would function in the dead time of local ASR(s) in order to be compatible with ATCRBS. It is expected that this method of operation will be technically feasible unless high surface interrogation rates exist due to radiation from surrounding ATCRBS Analytical studies have indicated that the surface interrogation environments at Chicago O'Hare and Los Angeles International Airports might be very severe. It was necessary, therefore, to measure exactly the surface interrogation rates at these airports since both are candidates for advanced surface surveillance systems.

The test program was carried out by the following team:

- H. R. Jackson, Federal Aviation Administration-National Aviation Facilities Experimental Center
- J. D. Vinatieri, The Mitre Corporation
- P. J. Woodall, Bendix Communications Division.

Our endeavors were greatly facilitated by cooperation from the Federal Aviation Administration Administration, the O'Hare Airport Administration Office and the Los Angeles Airport Management Office. In particular we wish to acknowledge the interest, encouragement and efforts of Mr. Jim Burns and Mr. Norm Oleson, FAA, Chicago O'Hare Airport; Mr. Jim Donovan, O'Hare Airport Administration Office; Mr. Frank Scollick, Mr. Doug LePage and Mr. Bob Curtis, FAA Los Angeles International Airport.

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CONTENTS

Section		Page
1. INTRO	DUCTION	1
1.2	PURPOSE OF THE MEASUREMENT PROGRAM PREVIOUS TESTS TEST OBJECTIVES	1 1 2
2. DATA	COLLECTION	3
2.2	METHOD OF APPROACH TEST EQUIPMENT SURFACE TEST PROCEDURES	3 3 5
3. DATA	ANALYSIS	9
3.2	GENERAL APPROACH SURFACE INTERROGATION ANALYSIS 3.2.1 Chicago O'Hare Data 3.2.2 Los Angeles International Data	9 10 10 14
	COMPOSITE DATA SUMMARY AVAILABILITY OF TIME FOR ASTC USE	20 20
4. CONCL	USIONS	23
APPENDIX A:	SPECIFICATIONS FOR PORTABLE INTERROGATION AND SLS MEASUREMENT EQUIPMENT	25
APPENDIX B:	CHICAGO O'HARE DATA	29
APPENDIX C:	LOS ANGELES DATA	31
PEFFRENCES		3/

ILLUSTRATIONS

Figure		Page
2-1	Packaging of Equipment Showing Storage of Antenna Ground Plane and All Necessary Cables	4
2-2	Equipment Shown With Transponder Antenna and Ground Plane Removed	6
3-1	Surface Interrogation Map for Chicago O'Hare Airport	11
3-2	Typical Interrogation Histogram for O'Hare Airport	13
3-3	Interrogation Histogram For O'Hare Airport Showing Burst from Two Interrogators	15
3-4	Surface Interrogation Map for Los Angeles International Airport	16
3-5	Typical Interrogation Histogram for Los Angeles International Airport	18
3-6	Interrogation Histogram for Los Angeles International Airport Showing Phantom Interrogations	19
A-1	Portable Interrogation and SLS Measuring Equipment	28
	TABLES	
Table		Page
3-1	Surface Interrogation Data Summary	21

LIST OF ABBREVIATIONS AND SYMBOLS

ARSR Air Route Surveillance Radar

ARTCC Air Route Traffic Control Center

ASR Airport Surveillance Radar

ASTC Airport Surface Traffic Control

ATC Air Traffic Control

ATCRBS Air Traffic Control Radar Beacon System

BCN Beacon

DAS Data Acquisition Subsystem

IFR Instrument Flight Rules

ISLS Interrogate Side Lobe Suppression

1²SLS Improved Interrogate Side Lobe Suppression

NAFEC National Aviation Facilities Experimental Center

NAS National Airspace System

NM Nautical Mile

PRF Pulse Repetition Frequency

RSLS Receive Side Lobe Suppression

TRACON Terminal Radar Control

TSC Transportation Systems Center

SLS Side Lobe Suppression

VFR Visual Flight Rules

그러, 1분 등 기 공항

1.2

1. INTRODUCTION

1.1 PURPOSE OF THE MEASUREMENT PROGRAM

The purpose of measuring the Surface Interrogation Environment at Chicago O'Hare (ORD) and at Los Angeles Airport (LAX) was to enlarge the body of data previously collected at the NAFEC and Logan Airports. The O'Hare and Los Angeles airports were selected because they are representative of large hubs and are candidates for installation of an operational ATCRBS-based Airport Surface Traffic Control (ASTC) system. Furthermore, O'Hare Airport is being considered for advanced testing of the brassboard model ATCRBS-based Surface Trilateration Data Acquisition Subsystem presently being built by TSC for planned feasibility tests at NAFEC. O'Hare also is a candidate site for installation and checkout of the Tower Automated Ground Surveillance (TAGS) system which will be the advanced ASTC system based on ATCRBS-trilateration as the primary surveillance sensor. Hence, there was a need to measure the interference environment at O'Hare. On the other hand, it has been suggested that the severity of the surface interrogation environment at Los Angeles might preclude the operation of an ATCRBS-based ASTC system. Indeed, analytical studies indicated that as many as ten interrogators might have lineof-sight coverage of the Los Angeles airport surface. Measurements at this site were needed to validate the analytical results and to provide valuable data representative of a heavy surface interrogation environment. The tests at both airports were performed in accordance with the test plan of reference 1.

1.2 PREVIOUS TESTS

The tests performed at these airports were similar to those previously made at Logan Airport and reported in Reference 2. Measurements were made using portable equipment which could easily be transported by automobile to various locations on the airport surface. This equipment was used to gather information on interrogation and improved interrogate side lobe suppression (I²SLS)* rates occurring for a surface transponder at designated airport locations.

^{*}ISLS generally refers to Interrogate Side Lobe Suppression whereby the P_2 pulse is radiated from an omni antenna. The FAA has widely implemented a system known as "improved ISLS" (I^2SLS). In this system, a portion of the P_1 signal is radiated along with P_2 through the omni antenna to suppress all transponders within the omni coverage volume not in the mainbeam. The equipment used in this test program does not differentiate between the two types, but simply records all suppression events occurring in the transponder whether from ISLS or I^2SLS. Henceforth, in this document, the transponder suppression rate will be referred to as the I^2 SLS suppression rate. Reference 1, pages 2 and 3, describe both suppression systems in more detail.

1.3 TEST OBJECTIVES

The objectives for this test program were: (1) to collect data at ORD and LAX airports and (2) perform a detailed analysis of the data in order to: (a) provide a first order assessment of the extent to which a surface operated transponder is interrogated by existing ATCRBS interrogators; (b) determine the uniformity of the surface transponder interrogation rate as a function of airport surface location; and (c) determine the number of interrogators in the vicinity of each airport causing surface transponder interrogations.

The field test data were obtained from a specially constructed test unit containing an instrumented transponder and associated electronics. This equipment provided data relative to the surface interrogation rate and $\rm I^2SLS$ rate and, through post-test analysis, provided a measure of the number of interrogators illuminating each test point.

Specifically, the test objectives were as follows:

- a. To take multiple counts of the number of interrogations and suppressions occurring during specified time intervals at several airport locations. These counts were later compared and analyzed.
- b. To determine the periodicity of bursts of interrogations occurring during specified time intervals,
 and to compare these periodicities with known beacon
 interrogator scan rates. By means of this procedure
 the number of interrogators illuminating a given test
 location was determined.

2. DATA COLLECTION

2.1 METHOD OF APPROACH

As previously stated, tests were made at the Chicago O'Hare and Los Angeles airports. In general, the same three basic steps were used at each airport site. First, a preliminary meeting was held at which the test team, composed of representatives from TSC, FAA, MITRE and the Bendix Communications Division, were introduced to FAA and Airport Authority operating personnel. A briefing was given by TSC to explain the purpose of the visit, the objectives of the ASTC program at TSC and the relationship of data that would be collected at the airports to the overall program. At this meeting, the manner in which data would be gathered on the airport surface was stressed and the interface between the test team, ATC controllers and Airport Authority personnel was discussed in detail. Also at this meeting, the operating characteristics (e.g., scan rate, PRF, stagger) of the local ASR beacon(s) were reviewed and the location and identity of the ASR and ARSR beacons in the airport vicinity which might have line of sight to the airport surface were noted.

Second, the test data was collected over a period of one to one and one-half days by driving an instrumented transponder to the planned test points. The test team and test equipment were driven on the airport surface by an authorized driver in a vehicle supplied by the FAA or the Airport Authority.

Third, a departure briefing was held at the conclusion of the test data gathering. Highlights of the testing were discussed and an attempt was made to give a very preliminary assessment of the interrogator environment based on a cursory examination of the data. It was emphasized that the data analysis would take several weeks and the test results for both airports would be presented in a final report.

2.2 TEST EQUIPMENT

At both airports portable equipment (see Figure 2-1) was employed to obtain measurements at selected locations. A major component of this equipment was a transponder (see Appendix A) which had been modified to provide outputs from its decoder circuitry in order to allow counts to be made of: (1) the number of valid interrogations received and (2) the number of suppressions which occurred during a prescribed time interval. The time interval was selectable in 0.1 second increments from 0.1 seconds to 10.0 seconds and the equipment had provision for recording in sequence the counts for 61 such intervals. At the conclusion of 61 intervals, the total count was printed and the counter automatically reset in preparation for the next test sequence.

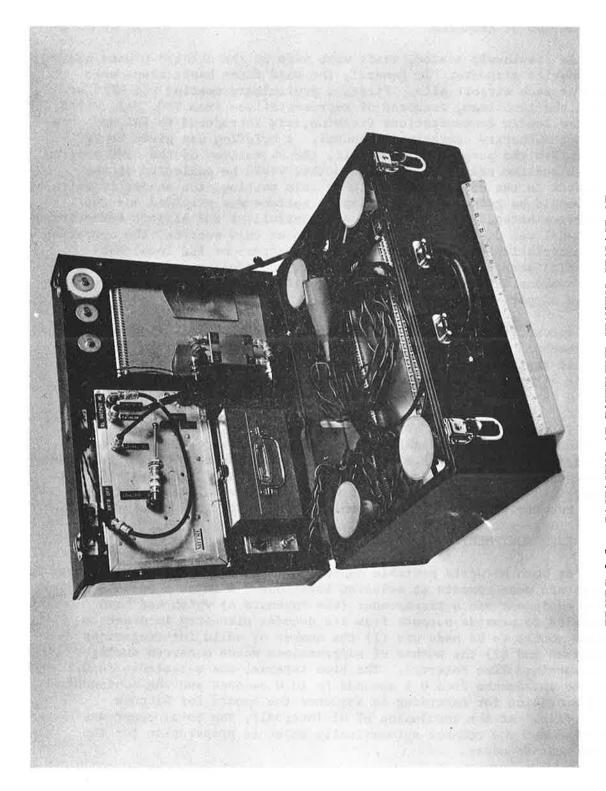


FIGURE 2-1. PACKAGING OF EQUIPMENT SHOWING STORAGE OF ANTENNA GROUND PLANE AND ALL NECESSARY CABLES

The equipment was designed so that it could be easily installed and operated in an ordinary surface vehicle such as an automobile or truck for rapid transportation to each test point. Set up time in the vehicle required only 5 to 10 minutes. A stub antenna was mounted on an aluminum ground plane which contained four suction cups. This antenna assembly was placed on the roof of the vehicle and held in place by two clamps fastened to the vehicle's rain gutter. Power for the equipment was provided by the 12 volt car battery either through battery cables and clips or by means of a plug inserted into the cigarette lighter.

Figures 2-1 and 2-2 illustrate the packaging of the equipment. Figure 2-1 shows the equipment with the necessary cables, antenna and ground plane packaged in a single case. The transponder can be seen mounted in the lid of the case and the counter electronics and strip printer are shown directly beneath the antenna ground plane. Figure 2-2 shows the equipment as it is set up at an airport site with the exception, of course, that the antenna and ground plan will be fastened to the roof of a vehicle. The counter electronics can be clearly seen and the strip printer is shown immediately to the left of the electronic assembly.

Before starting the series of measurements at O'Hare and Los Angeles, the transponder was bench tested at MITRE. This bench test showed the transponder to have a sensitivity of -72 dbm and a power output of 220 watts. These values are within the ATCRBS national standard* and thus the interrogation and I²SLS counts obtained are representative of typical operational values. In addition, the portable equipment was checked out at Logan airport prior to the initiation of these tests. This field check provided verification that the transponder and counter electronics were properly interfaced and were accurately recording the surface interrogation rate.

2.3 SURFACE TEST PROCEDURES

Testing was accomplished using the portable test equipment to record the number of successful interrogations occurring in a specified time interval and the number of successful sidelobe suppressions received by the transponder during a specified time interval. After the equipment was installed in the test vehicle a check was made to verify its readiness. Once installed, the equipment, including the antenna fastened to the roof and the battery

^{*}U.S. National Standard for the IFF Mark X Air Traffic Control Radar Beacon System Characteristics (ATCRBS), AC No. 00-27, Jan. 1969

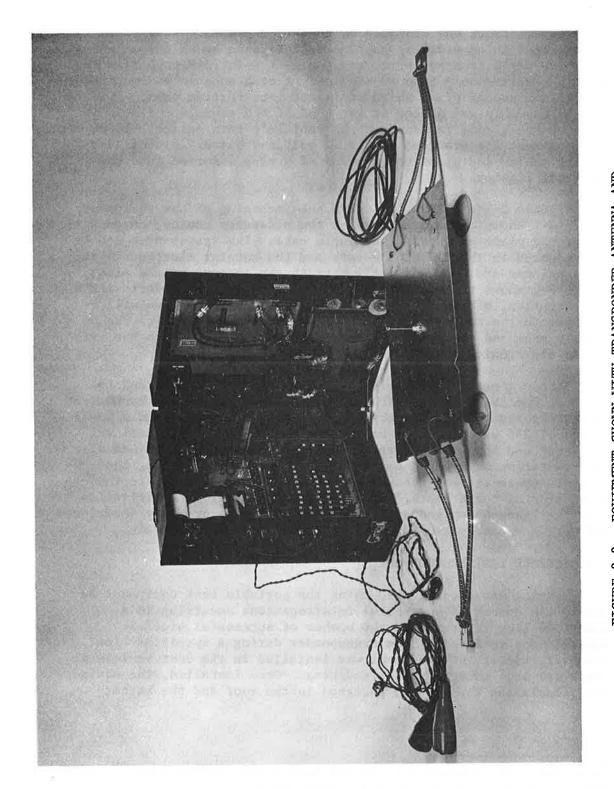


FIGURE 2-2. EQUIPMENT SHOWN WITH TRANSPONDER ANTENNA AND GROUND PLANE REMOVED

connection, remained in place while the vehicle travelled between test locations. In this way a minimum amount of time was required at each test point since the equipment was ready for data collection when the vehicle arrived at each test point. At both airports, the vehicle was driven by an experienced driver familiar with the airport runway and taxiway system. In addition the test team had VHF communications with ground controllers to request and receive clearances when needed to proceed to specific test points.

The surface interrogation and suppression rate due to each interrogator illuminating the airport surface was measured by taking interrogation counts over time intervals of 0.5 second and sidelobe suppression counts over time intervals of 1.0 seconds. Using this procedure, interrogation counts were made in a 30.5 second period based on data collected for 61 intervals. In general, Airport Surveillance Radar (ASR) interrogators have a scan rate in the order of 4 seconds and Air Route Surveillance Radar (ARSR) interrogators have a scan rate of about 12 seconds. A 30.5 second sample therefore provided for roughly 7 scans of ASR beacon data and 2-3 scans of ARSR data. During the post-test analysis, the strip printer tapes containing 0.5 second counts for each of the 61 intervals recorded were examined for periodicity of interrogation counts to determine the number of ASR and ARSR beacon interrogators illuminating the field.

The I²SLS counts measured over time intervals of 1.0 second provided data relative to the suppression rates of surface transponders, and were correlated with interrogation counts and known PRFs for interrogators in the vicinity of the airport. This provided an additional aid in evaluating the surface interrogation environment.

Total counts of the number of interrogation and I²SLS events during the 30.5 second data collection period also were printed on the paper tapes. In addition, another sample was collected using a 1.0 second sampling rate to measure total suppression and interrogation counts. Light emitting diodes (LED) were used to indicate total counts over the 61 second test period rather than the paper tape readout.

Although original test points were located directly on runway and taxiway surfaces and would have required the test vehicle to be parked at these locations for periods of five to ten minutes, interference with normal airport surface operations was effectively eliminated by parking the vehicle on access roads and grass areas adjacent to test points. Thus test points in Figures 3-1 and 3-4 represent only approximate locations where measurements were made. In this manner, the performance of these tests did not obstruct the normal flow of surface traffic and conversely surface traffic did not impede the collection of data.

Data Recording Requirements

In the test vehicle, a record (log) of all significant events which occurred during the test was maintained. This log included the following types of data:

- Test point number
- Date
- Time-of-day
- Run number
- Average interrogation and I²SLS counts for each run
- Comments such as equipment malfunctions and difficulties observed.

It also was a required procedure that each paper tape produced by the test equipment was annotated with the following data:

- Test point number
- . Time-of-day
- Interrogation or I²SLS data
- . Time interval used to collect data
- Run number at each location.

3. DATA ANALYSIS

3.1 GENERAL APPROACH

The equipment used to collect the data was different than that used in earlier tests (see References 2 and 3), but provided the means for collecting data with greater precision than that previously obtained. These data included not only the average surface interrogation rates similar to those obtained at Logan and Atlantic City airports but also individual bursts of interrogations as each radar scanned the airport surface. These data were useful in determining how many interrogators actually illuminated the airport surface. Individual interrogation bursts were analyzed to assess the effective interrogation beamwidth of individual interrogators. This provided a means to discriminate between interrogations from ASR beacons and ARSR beacons. Bursts also were examined for periodicities equal to scan rates of nearby interrogators. This helped to develop an approximation of the number of interrogators with line-of-sight to the airport surface. This estimate was substantiated by correlating the SLS total count with the sum of the PRF's for nearby interrogators.

Once the analysis had established the number of interrogators illuminating the surface, the test data was compared with analytical data. Analytical results on the number of interrogators illuminating each airport surface had been derived by using the known locations of interrogators in the vicinity of each airport and by examining topographical maps to determine those interrogators with probable line-of-sight coverage of the airport surface. This method was suspect because factors such as the limited information on the topographical maps, their date, and unknowns such as recent construction around the airport perimeter which could block or shield the surface from interrogations. In the analysis, peripheral interrogators were assumed to contribute to the surface transponder interrogation rate. The results from this test series established a measure of the validity of predicting surface interrogation rates for any airport based on anything other than a measurement of the surface environment.

Data which have been collected at Logan and Atlantic City Airports (NAFEC) were designed to assess the extent of the interrogation and side lobe suppression of surface vehicle transponders at various airport locations. NAFEC offered the opportunity to examine these effects under various conditions of interrogator population ranging from one to three beacon interrogators. In addition, data were collected during the NAFEC test series for various configurations of ASR and ARSR beacon interrogators. Logan Airport, on the other hand, offered the opportunity for examination of surface interrogation rates at higher surface traffic volumes.

One result of the previous tests was the generation of several interrogation maps of each airport surface showing the surface vehicle interrogation rates at various locations on the airport surface. Since the O'Hare and Los Angeles tests were designed to extend these results and thereby increase the data base, the final step in the analysis was a comparison of data obtained at O'Hare and Los Angeles with data previously obtained at Logan and NAFEC.

3.2 SURFACE INTERROGATION ANALYSIS

Data collection consisted of counting the number of interrogation and SLS events occurring in the instrumented transponder. As previously explained, the types of data collected at each test location were:

- a. Interrogation counts at 0.5 seconds using paper tape
- b. Total interrogations for 30.5 seconds using paper tape
- c. Total I²SLS counts for 30.5 seconds using paper tape.
- d. Total interrogations for 61 seconds using LED
- e. Total I 2 SLS counts for 61 seconds using LED.

3.2.1 Chicago O'Hare Data

3.2.1.1 Surface Interrogation and SLS Map. Interrogations of the surface transponder occurred in bursts as the O'Hare ASR interrogator scanned the airport. These data have been broken down into the average number of replies per ASR scan and the average replies per second. An interrogation map for the 25 test points showing the replies per second and SLS counts per second is presented in Figure 3-1. Two numbers are shown at each test point. The top number is the number of successful interrogations (replies generated) per second. The bottom number is the number of SLS counts generated per second.

From Figure 3-1 it can be seen that the average surface interrogation rate at O'Hare is very low; generally ranging between 7 and 13 counts per second. This indicates that the only source of surface interrogations was the local O'Hare ASR interrogator. This can be confirmed by observing that the SLS counts which show an average of 408 per second is very close to the average O'Hare PRF of 423 per second. Moreover, the sum of the average interrogation rate (12.8) and the average SLS rate (408) is seen to be 421 per second versus the known PRF of 423 per second. This provides a better than 98% correlation. A comparison of the O'Hare data with Logan Data reveals that although O'Hare has only a single ASR interrogator, as opposed to



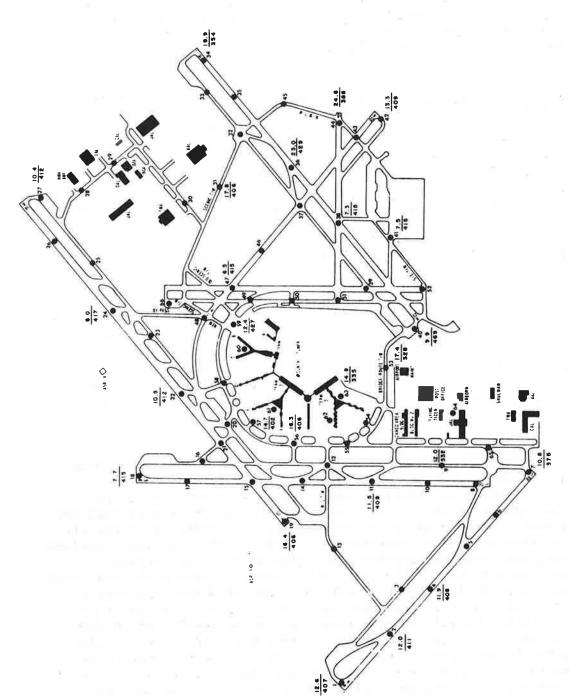
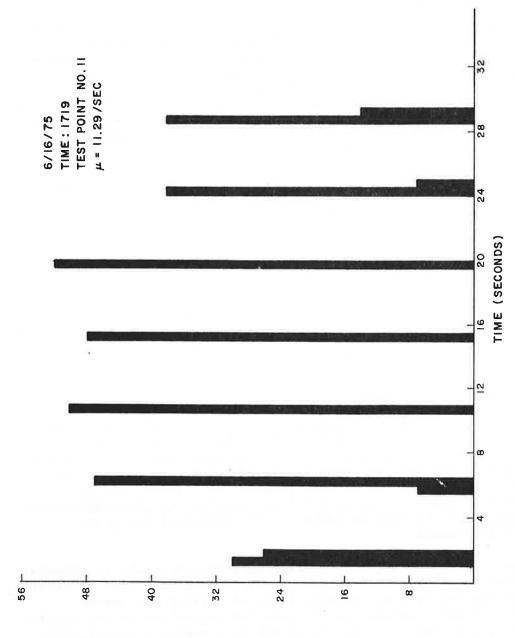


Figure 3-1 SURFACE INTERROGATION MAP FOR CHICAGO O'HARE AIRPORT

the single ASR single ARSR interrogator environment at Logan, the average reply rate for O'Hare is 13, compared to 9.5 for Logan. This can be explained by the fact that the average number of replies per scan triggered by the O'Hare ASR interrogator is somewhat higher than that triggered by the Logan ASR interrogator suggesting that the beamwidth of the O'Hare interrogator is greater than that at Logan. Nevertheless, the reply rate at O'Hare is consistently at a very low value.

- 3.2.1.2 Effective Interrogation Beamwidth. Figure 3-2 shows a typical example of the temporal distribution of replies triggered from the surface transponder. Replies are seen to occur in bursts of 40 to 50 at periodic intervals of approximately 4.5 seconds. This coincides very well with the O'Hare scan rate of 4.62 seconds. The average number of replies per scan is 40 which is considerably larger than the 18.5 per scan previously measured at Logan. The differences between these two values must be regarded with some reservations because the reply per scan data for Logan was inferred based upon total reply counts measured over 70 scans whereas a high sampling rate approach was employed at O'Hare. Nevertheless, the average reply rate has been shown to be slightly higher for O'Hare. Hence there is a great likelihood that the reply per scan counts are, in fact, higher for O'Hare than Logan.
- 3.2.1.3 Number of Interrogators. Examination of Figure 3-2 clearly shows that only one interrogator is triggering replies since only one set of interrogation bursts is present with a periodic rate of 4.5 seconds and essentially no interrogations are present between these bursts. Thus, it can be concluded that only the local O'Hare ASR interrogator is successful in triggering surface transponder replies.

In the previous analytical study which was performed to provide data on the number of interrogators within line-of-sight to the O'Hare airport surface, vertical contour maps were constructed for each interrogator in the vicinity of the airport. Those interrogators with an unobstructed view (based on available information) of the airport were then assumed to contribute to surface interrogation, provided, of course, that the interrogator was active at the time surface interrogation data were collected. The results of this study indicated that there were 12 interrogators within 25 nmi of O'Hare and that of these 12 interrogators, 8 probably had line-of-sight coverage of the O'Hare surface. This is not substantiated by the results which show that only one interrogator actively interrogates surface transponders. It is possible that some of the 8 interrogators were not operational while surface measurements were being made. However, as was indicated in Section II, test data were collected on two successive days and for several time periods. Thus, it seems reasonable to conclude that



TYPICAL INTERROGATION HISTOGRAM FOR O'HARE AIRPORT Figure 3-2

INTERROGATION COUNTS

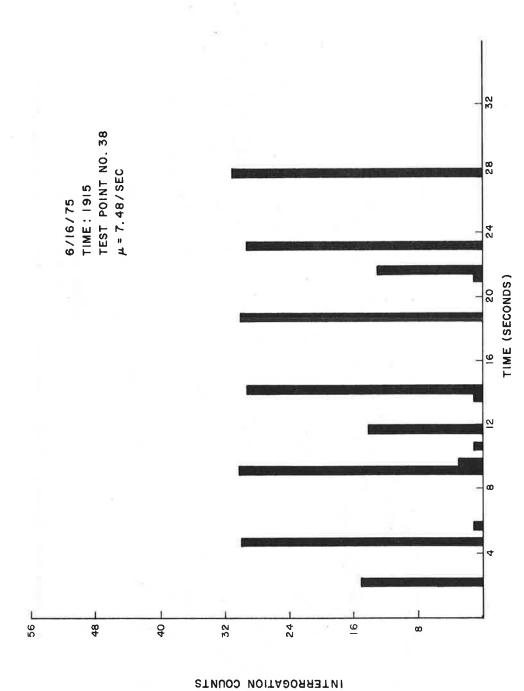
the nominal surface interrogation environment at O'Hare is composed of a single ASR interrogator. Furthermore, based on these results, the use of topographical maps and known interrogator locations as a means for estimating the surface interrogation environment at a given airport does not appear to be reliable. In fact, the examination of topographical maps indicated that the McCook ARSR located approximately 10.5 nmi southwest of the airport would have line-of-sight coverage of the airport surface and thus contribute to the surface interrogation environment. Since this interrogator was known to be operational at all times, but was not observed in the test data, it must be concluded that the interrogator does not radiate energy to the airport surface. Once again, this can be accounted for by the many additional variables such as man-made obstructions, antenna patterns, etc., that were not taken into account in the analytical studies.

Finally, on one occasion there seemed to be a second interrogator triggering surface replies. These data are shown in Figure 3-3. In the histogram, the O'Hare ASR interrogations are still present at a 4.5 second periodicity, but now additional replies are generated at a periodicity of approximately 10 seconds. These interrogations may have come from the McCook ARSR interrogator or possibly from a nearby military installation. In any event these additional interrogations were observed only on this one occasion.

3.2.2 Los Angeles International Data

3.2.2.1 Surface Interrogation and SLS Map. The same analysis technique previously described in Section 3.2.1 for the O'Hare data was also used for the Los Angeles data. An interrogation map of the replies per second and SLS gates generated per second is shown in Figure 3-4 for each of the 16 test points. Here again, the top number is the number of successful interrogations (replies generated) per second, while the bottom number is the number of SLS gates generated per second.

Figure 3-4 shows that the average interrogation rate is somewhat higher than O'Hare but, nevertheless, still relatively low. Typically the interrogation rate was between 20 to 30 replies per second. Subsequent analysis showed that surface interrogations at Los Angeles were caused principally by two local ASR interrogators and one ARSR interrogator. Although these three interrogators can be identified in the data, the SLS rate is only 767 per second, which is roughly equivalent to the sum of the average PRF for both local ASR interrogators. Thus, the I²SLS of the ARSR interrogator is not actively suppressing surface transponders at Los Angeles even though this interrogator is triggering surface transponder replies.



INTERROGATION HISTOGRAM FOR O'HARE AIRPORT SHOWING BURST FROM TWO INTERROGATORS Figure 3-3

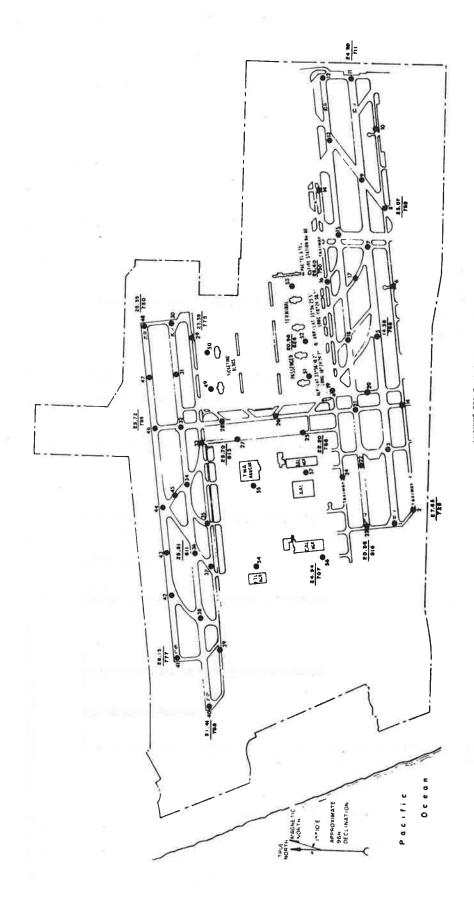


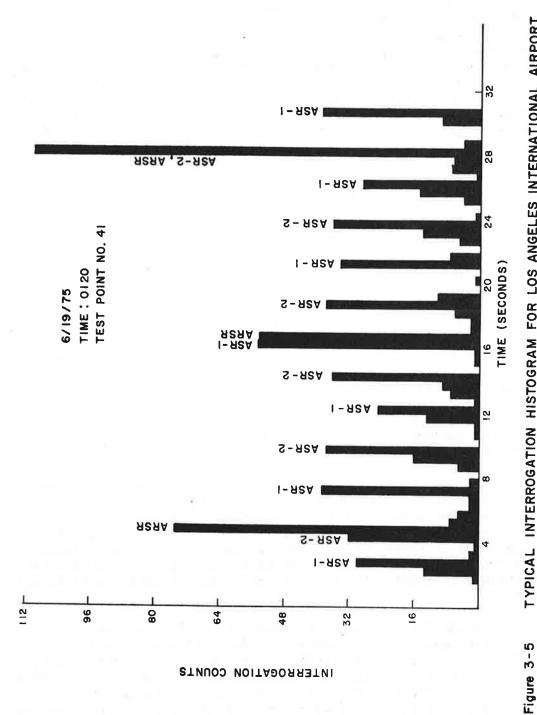
Figure 3-4 SURFACE INTERROGATION MAP FOR LOS ANGELES INTERNATIONAL AIRPORT

3.2.2.2 Effective Interrogation Beamwidth. Figure 3-5 shows a typical example of the temporal distribution of replies triggered from the surface transponder at LAX. Three sets of interrogation bursts are evident. Two sets have periodicities of roughly 4.8 seconds corresponding to the scan rates of the Los Angeles ASR-7 and ASR-4 beacon interrogators. It was not possible with this test data to determine the particular interrogator responsible for each set of interrogation bursts. The third set has a periodicity of 12 seconds and is in good agreement with the 12 second scan rate of the ARSR interrogator located approximately 5 miles southwest of the airport.

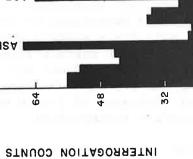
Examination of the number of interrogations in each burst reveals that for both ASR interrogators there are about 32 interrogations per scan. This is somewhat less than the O'Hare ASR interrogator which triggered around 48 replies per scan, but it is more than the 18.5 per scan generated by the Logan ASR interrogator. The ARSR interrogator caused 70 to 80 replies per scan and seems to be consistent with the beamwidth and scan rate for this radar. Between interrogation bursts the environment was relatively clean. The data presented in Figure 3-5 was collected at 0120 local time. It is interesting to observe that data collected at the same test location at 1123 the following morning show nearly identical results. Nevertheless, the distribution of data from test point to test point shows more variations than that previously experienced at either O'Hare or Logan. For example, the average reply rate was seen to vary from a high of 53 per second at test point 7 to a low of 19 at test point 14 (see Appendix C). This can be attributed to the numerous buildings on the airport surface which for various locations can cause shielding for either the ASR-7 or ASR-4 interrogators.

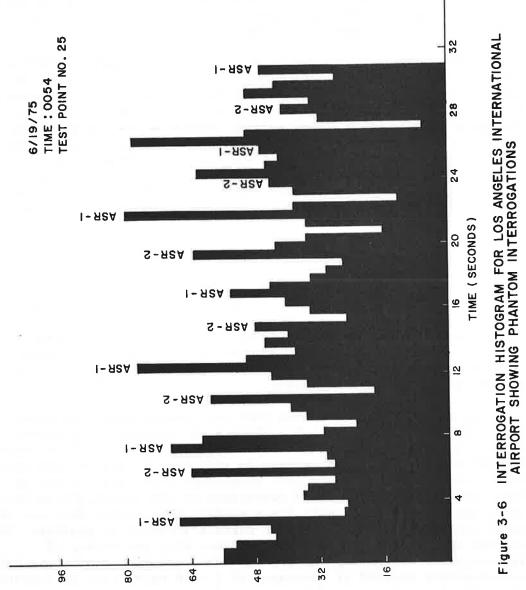
3.2.2.3 <u>Number of Interrogators</u>. The surface interrogation environment at Los Angeles as depicted in Figure 3-5 is due principally to three interrogators: 2 ASR interrogators and one ARSR interrogator. Analytical studies indicated that 19 interrogators are within 25 nmi of LAX, and of these, 10 had line-of-sight to the airport surface. This again points out the hazards in attempting to use analytical techniques to predict the surface interrogation environment at airports.

Although the nominal environment at Los Angeles was seen to be three interrogators, this was not always the case. The first test period at Los Angeles from 1600 to 2051 on June 16 provided data which indicated very high interrogation rates in the order of 80 to 90 per second and for every test point with the exception of test location number 8. A typical example of the data collected during this period is shown in Figure 3-6 for test location 25. It can be seen that interrogation bursts for the three radars are present,



TYPICAL INTERROGATION HISTOGRAM FOR LOS ANGELES INTERNATIONAL AIRPORT





however, the clean environment between the bursts is now filled in by additional interrogations. Moreover, the replies per scan now appear to be larger than those indicated by Figure 3-5. Thus, it appears that the entire data is biased by 20 to 30 additional interrogations per second. This phenomenon could be caused by some form of omni interrogation signal operating at this low PRF.

At the outset of the second test period from 0020 to 0300 on June 19, the same omni interrogation effect was encountered. At 0100, however, the environment immediately changed; the omni interrogations disappeared and data similar to that shown in Figure 3-5 were obtained for the remainder of the test period. These values are shown in Appendix C. The dramatic change in the environment suggests that precisely at 0100 an interrogator was turned off thereby reducing the average reply rate from 80 per second to 24 per second. Time did not permit identification or location of the omni interrogator, however, it did remain off during the third and final test period on the morning of June 19.

3.3 COMPOSITE DATA SUMMARY

Table 3-1 provides a summary of the interrogation and SLS rates measured for the Logan, NAFEC, O'Hare, and Los Angeles airports. In each case the number of ASR and ARSR interrogators triggering surface transponder replies has been indicated. The average interrogation and SLS rates for each airport are listed with their respective standard deviations. These data were obtained using values measured at each airport's test locations.

The data in Table 3-1 show that the interrogation environments even for Los Angeles, which has often been suggested as a severe surface interrogation environment, are in fact very low when compared to airborne interrogation rates. In addition, it can be seen that the interrogation rates are very consistent over each airport as indicated by the low standard deviations of the interrogation rates. Thus, the interrogation rate for a surface transponder does not show large variations from one airport surface location to another. The average SLS rate is seen to be consistent with the number of interrogators triggering replies with the exception that those ARSR interrogators located at distances of 3 or 4 miles from an airport do not actively suppress surface transponders.

3.4 AVAILABILITY OF TIME FOR ASTC USE

To prevent mutual interference between ATCRBS and ASTC systems, the ASTC system is being designed to synchronize ASTC interrogations to occur only in the dead-time of the local ATCRBS equipment. Should

TABLE 3-1

SURFACE INTERROGATION DATA SUMMARY

AIRPORT	DATE	NUMBER OF TEST POINTS	NUMBER OF INTERROGATORS	R OF GATORS	AVG. INTERR. RATE PER SEC.	STD. DEV. OF INTERR. RATE	AVG, SLS RATE PER SEC,	STD. DEV. OF SLS RATE
			ASR	ARSR				
Logan	Dec. 1973	52	1	1	05*6	2.69	777	35
NAFEC	March 1974	45	2	П	13.18	2,53	753	٠,
0'Hare	June 1975	25	н	0	13.12	4.87	408	38
Los Angeles	June 1975	16	2	-	24.21	8.83	767	36

synchronization be required with additional sites, the available time for active ASTC use would be reduced resulting in a possible reduction in the ASTC update rate. As can be seen from Table 3-1, only one interrogator illuminates the O'Hare airport surface and as such synchronization would be required only to the local O'Hare ASR beacon interrogator. This synchronization would not affect the ASTC update since the system is being designed with this requirement in mind.

The environment at Los Angeles is somewhat more severe than that at O'Hare since there are 2 ASR interrogators and one ARSR interrogator illuminating the surface. It is unlikely that synchronization will be required with the ARSR since mutual interference with this site should be low due to its relatively remote location. Nevertheless there remains 2 ASR interrogators to which the ASTC system may have to be synchronized. If this is the case, and assuming the 2 ASRs are not synchronized to each other but are running at approximately the same PRF, the available time for ASTC use may be reduced by as much as one-half. This should not severely impact system performance, however, because sufficient margin has been allowed to provide necessary update rates even for this reduced ASTC PRF requirement. For example, the ASTC system is designed to provide a maximum of 14 interrogations per ATCRBS dead-time. If it is required to synchronize to a second ASR, only 7 ASTC interrogations may be available per equivalent ATCRBS dead-time. Initial estimates suggest, however, that only an average of 3 or 4 ASTC interrogations per dead time may actually be necessary to provide the required update rates. It therefore does not appear that the presence of 2 ASR interrogators at Los Angeles will in any way impact on the implementation of an ATCRBSbased ASTC system. Furthermore, it is not certain at this time whether it will even be necessary to synchronize to both radars.

4. CONCLUSIONS

The analysis and data presented in this report support the following specific statements concerning the surface interrogation environment at Chicago O'Hare and Los Angeles airports.

- a. Results show only a single ASR interrogator at O'Hare triggering surface transponder replies. The average surface interrogation rate at O'Hare is 12.82 per second which is only slightly higher than the 9.5 per second measured at Logan airport. It is recommended that surface interrogation data be re-measured at Logan using the new portable equipment.
- b. Results show that two ASR and one ARSR interrogators trigger surface transponder replies at Los Angeles. The average surface interrogation rate at Los Angeles is 24.21 per second which is considerably higher than Logan and O'Hare but, nevertheless, is much less than either initial expectations or typical airborne interrogation rates.
- c. Although the nominal environment at Los Angeles consists of three interrogators as identified by 2 above, additional surface interrogations can occur from an unknown source. Extraneous interrogations were present during a portion of the tests at Los Angeles and appeared to be from an omnidirectional interrogator using a PRF between 30 and 40 per second.
- d. Interrogation rates at both Chicago O'Hare and Los Angeles are very uniform for the respective airport surfaces, with little variation among the data collected for various surface locations.
- e. Analysis of data indicates that sufficient time will be available for ASTC use at both Chicago O'Hare and Los Angeles airports. The ASTC system will have to be synchronized to only one ASR interrogator at O'Hare and at most two ASR interrogators at Los Angeles.
- f. Data obtained from these tests have shown that the use of topographical maps is not a reliable technique for estimating the surface transponder interrogation environment at airports since analytical results can grossly overestimate the interrogation environment. These tests have clearly demonstrated the utility of field test data in making such determinations.

APPENDIX A

SPECIFICATIONS FOR PORTABLE INTERROGATION AND SLS MEASUREMENT EQUIPMENT

COUNTER ELECTRONICS

INPUT

Switch selectable for either interrogation or SLS suppression counts.

OUTPUT

Interrogation or SLS counts (depends on input switch selector setting) occurring during each sample interval. Data are output for each of 61 sample intervals. Total count is output in interval 62 and total SLS or interrogation count is output in interval 63. For example, if interrogation counts are output in each of the 61 intervals, then total interrogation count will appear in interval 62 and total SLS count will appear in interval 63 and vice versa.

OPERATOR FEATURES

- Start Manual by push button action
- Stop Automatic at end of 61 sample intervals
- Reset (a) Automatic at end of 61 sample intervals
 (b) Manual by push button action at any time
 during data collection period
- Readout (a) Light emitting diode (LED) readout
 - (b) On-line paper tape printout

SAMPLE INTERVAL TIME PERIOD

Switch selectable in increments of 0.1 seconds from 0.1 seconds to 10.0 seconds.

DATA COLLECTION PERIOD

Dependent on sample interval selected. This period can range from a minimum of 6.1 seconds to a maximum of 610 seconds.

MEMORY

63 registers each containing 4 decimal digits used for storing data for LED readout at the completion of each data collection period.

PRINTER DISPLAY

- a. On-line printouts of counts measured in each of the 61 sample intervals, plus total count in intervals 62 and 63.
- b. Six column printout; four data columns plus two columns for interval number. Each column is one decimal digit.
- c. Printer usable with sample interval time periods greater than 0.4 seconds. Printer not usable below 0.4 seconds due to speed limitations.

READOUT DISPLAY

- a. Four digit LED readout for data plus two digit LED readout for memory location.
- b. Readout of each of the 61 memory locations by sequence using push button.
- c. Readout of memory location 62 by single push button action.

POWER REQUIREMENTS

12 volt supply with connection to either the vehicle cigarette lighter or directly to the battery.

ANTENNA

Stub type mounted on aluminum ground plane. Antenna assembly fastens to the roof of vehicle by two straps which are clipped over the rain gutter.

PACKAGING

Case measuring 18 inches by 12 inches by 8 inches.

TRANSPONDER

TYPE

KING KT-76

SENSITIVITY

-72 dbm

POWER OUTPUT

200 watts

POWER REQUIREMENTS

14 VDC, 1.3 Amps.

OUTPUTS

Interrogation - Single pulse output whenever a mode A or mode C interrogation is successfully decoded.

Sidelobe - Single pulse output whenever P₁ and P₂ pulses meeting suppression requirements are successfully decoded.

Transmitter - Switch selectable for either standby or operative. This function is independent of interrogation or SLS outputs.

Receiver - Output of transponder video available for viewing on an oscilloscope.

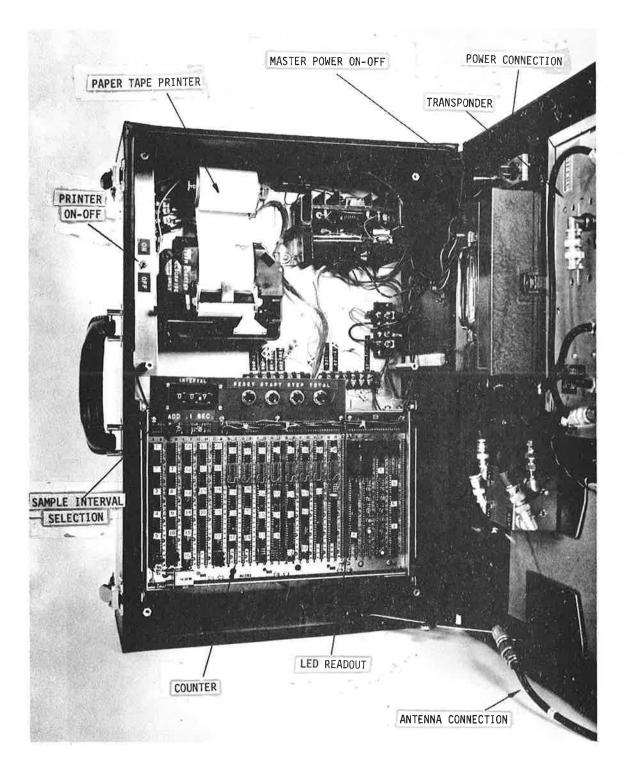


Figure A-1. Portable Interrogation and SLS Measuring Equipment

APPENDIX B
CHICAGO O'HARE DATA

TEST PERIOD I, JUNE 16

Test Point	Time-of- Day	Average Interr Rate per Second	Average SLS Rate per Second
1	1650	12.06	408
3	1703	13.33	405
4	1712	11.87	359
6	1721	11.29	411
8	1727	5.97	417
7	1740	16.39	406
9	1805	9.15	414
11	1816	10.78	412
13	1827	25.13	359
14	1857	15.93	375
15	1910	7.59	416
18	1920	15.35	416
16	1938	9.37	561
22	1958	9.45	413
21	2009	6.35	416
25	2020	14.07	406
24	2028	16.34	406
23	2033	14.99	335
4A	2045	11.64	485
		TEST PERIOD II, JUNE 1	<u>7</u>
1	0106	13.30	406
3	0115	10.61	412
4	0121	9.71	392
6	0128	11.68	407
7	0136	9.26	413

APPENDIX B (Concluded)

TEST PERIOD II, JUNE 17 (Concluded)

Test Point	Time-of- Day	Average Interr Rate per Second	Average SLS Rate per Second
8	0145	9.50	413
9	0150	12.03	411
11	0157	10.10	412
13	0203	14.73	349
14	0210	30.18	483
15	0217	7.08	415
18	0223	11.34	405
16	0225	10.45	378
21	0230	6.77	415
		TEST PERIOD III, JUNE 17	
10	1141	6.00	417
2	1145	12.00	411
4	1149	10.79	381
5	1156	12.00	532
7	1203	8.49	419
11	1215	9.48	413
12	1223	17.80	406
14	1228	44.62	402
17	1236	7.46	415
19	1243	24.61	386
20	1253	17.40	328
22	1307	12.47	427

APPENDIX C

TEST PERIOD I, JUNE 18

Test Point	Time-of- Day	Average Interr Rate per Second	Average SLS Rate per Second
3	1600	20.43	751
2	1610	35.98	976
1	1620	56.33	928
6	1625	94.20	758
7	1636	98.85	746
10	1646	27.61	746
11	1654	81.84	861
12	1702	88.36	726
13	1711	103.60	772
14	1720	89.13	655
5	1732	83.33	690
4	1740	105.92	729
			×
		TEST PERIOD II, JUNE 18	
3	1836	21.67	739
2	1843	52.61	747
1	1848	73.16	756
6	1942	90.72	713
16	1947	88.28	738
7	1950	91.74	781
8	1954	101.34	778
10	1958	20.07	754
11	2003	79.72	759
12	2007	78.62	791

APPENDIX C (Continued)

TEST PERIOD II, JUNE 18 (Continued)

Test Point	Time-of- Day	Average Interr Rate per Second	Average SLS Rate per Second
1.0	0010	. 7	
13	2013	84.10	754
14	2019	97.11	756
9	2024	24.54	780
15	2038	108.70	739
5	2043	98.11	747
4	2051	82.54	718
		TEST PERIOD III, JUNE 19	8 =
3	0020	37.31	704
2	0028	21.48	727
1	0037	86.57	786
6	0050	83.18	834
7	0054	83.25	776
8	0058	94.33	782
8A	0106	25.70	813
10	0109	26.39	962
11	0115	28.90	841
12	0120	27.27	862
13	0135	25.08	748
14	0140	21.57	755
9	0145	20.57	750
5	0152	26.12	748
15	0158	21.77	886
4	0205	25.67	677
3	0215	25.80	759
2	0218	8.90	762
1	0225	18.82	754

APPENDIX C (Concluded)
TEST PERIOD III, JUNE 19 (Concluded)

Test Point	Time-of- Day	Average Interr Rate per Second	Average SLS Rate per Second
6	0233	20.12	868
16	0236	22,52	710
7	0245	20.89	763
		TEST PERIOD IV, JUNE 19	
3	1041	20.34	757
2	1047	19.85	750
1	1052	36.47	704
6	1056	20.58	764
16	1101	29.36	705
7	1106	23.50	754
8	1110	54.48	725
10	1116	24.62	660
11	1119	33.91	735
12	1123	29.02	693
13	1130	22.35	762
14	1135	29.13	746
9	1140	26.21	801
15	1149	19.38	767
5	1157	24.68	752
4	1205	23.51	746

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